

# RF System Design

Peter Kinget

Bell Laboratories  
Lucent Technologies  
Murray Hill, NJ, USA

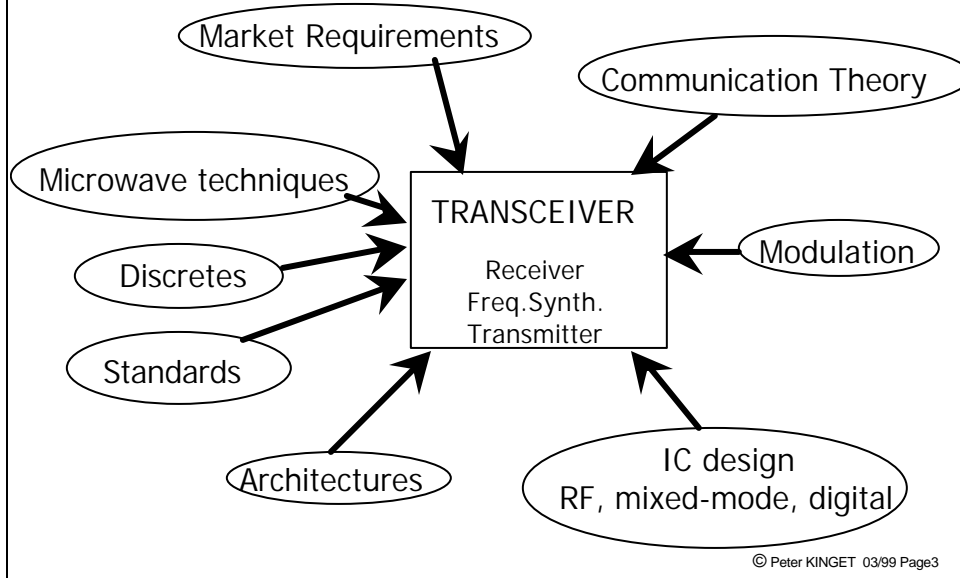
## Outline

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- Circuits for Wireless
- Wireless Communications
  - duplex, access, and cellular communication systems
  - standards
- Receivers:
  - heterodyne
  - homodyne
  - image reject
- Transmitters
  - modulation
  - up-conversion
- Transceivers
  - frequency synthesis
  - examples

## RF IC design

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## Circuits for Wireless

## Circuits for Wireless - Overview

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- Noise limits the *smallest* signal
  - noise figure
  - cascade of stages
- Distortion limits the *largest* signal
  - large (interfering) signals:
    - compression, blocking, and desensitization
    - inter-modulation
    - cascade of stages
- Dynamic Range

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## Noise Figure

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- Max. thermal noise power from linear passive network e.g. antenna:  $N_{\max} = kT \cdot BW$
- Noise Factor:  $F = \frac{(S/N)_{\text{in}}}{(S/N)_{\text{out}}} = 1 + \frac{N_{\text{eq@input}}^{\text{added}}}{N_{\text{in}}}$
- Noise Figure:  $NF = 10 \log_{10}(F) \geq 0\text{dB}$

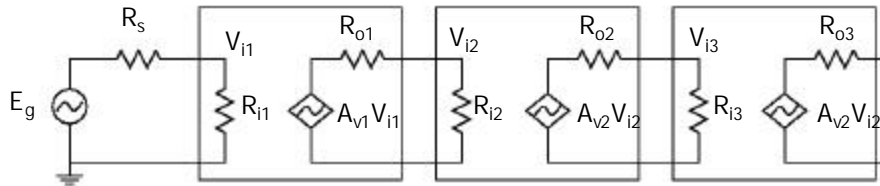
NF	F
[dB]	
0	1
1	1.25
2	1.6
3	2

$(S/N)_{\text{out}} = 1/2 (S/N)_{\text{in}}$

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## Cascade of Stages: Friis Equation

Avail. Power Gain:  $A_{p1}$        $A_{p2}$        $A_{p3}$   
 Noise factor:  $F_1$        $F_2$        $F_3$



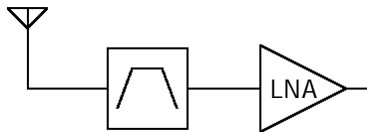
$$F = 1 + (F_1 - 1) + \frac{(F_2 - 1)}{A_{p1}} + \frac{(F_3 - 1)}{A_{p1}A_{p2}}$$

➔ later blocks contribute less to the noise figure if they are preceded with gain

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## Noise of lossy passive circuit

- Lossy passive circuit (e.g. filter):  $F = \text{Loss}$
- E.g. Band-select filter & LNA:



$$F = F_{\text{filt}} + \frac{(F_{\text{LNA}} - 1)}{A_p}$$

$$F = L + \frac{(F_{\text{LNA}} - 1)}{1/L} = L \cdot F_{\text{LNA}}$$

Loss adds immediately to noise figure !

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## Sensitivity

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- sensitivity = minimal signal level that receiver can detect for a given (S/N) at the output:

$$F = \frac{(S/N)_{in}}{(S/N)_{out}} = \frac{P_{signal\_in}}{P_{noise\_in}} \cdot \frac{1}{(S/N)_{out}}$$

$$\begin{aligned} P_{signal\_in} &= F \cdot (S/N)_{out} \cdot P_{noise\_in} \\ &= F \cdot (S/N)_{out} \cdot kT \cdot BW \end{aligned}$$

- E.g. GSM (BW=200kHz, (S/N)<sub>out</sub> > 9dB):

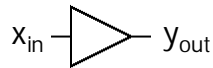
$$\begin{aligned} P_{signal\_in} &= NF + (S/N)_{out} - 174\text{dBm/Hz} + 10\log_{10}(BW) \\ &= 6 + 10 - 174 + 53 = -105\text{dBm} \end{aligned}$$

for a receiver with a noise figure of 6dB

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## Distortion:

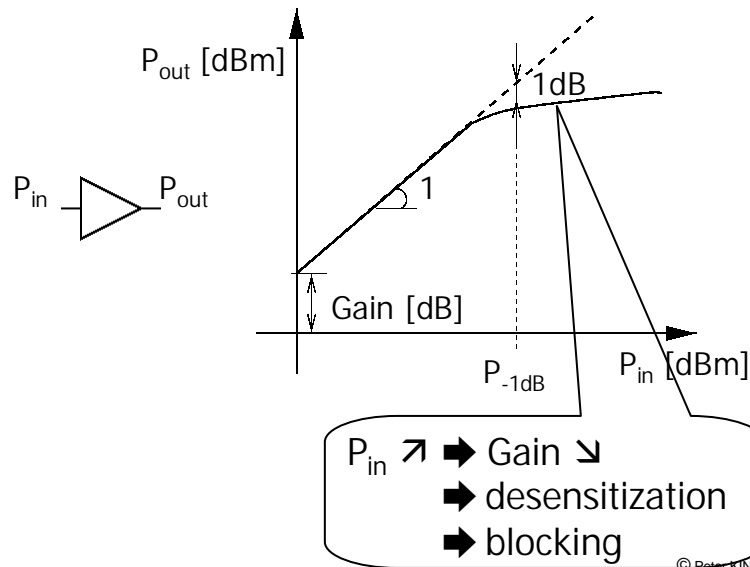
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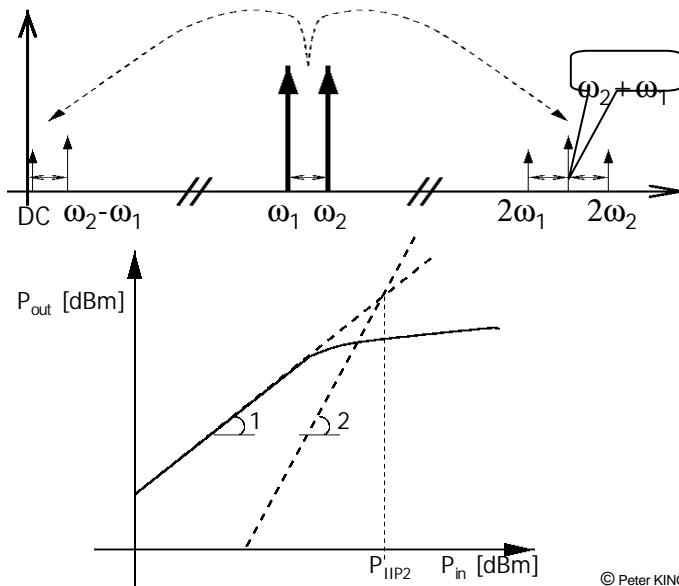
- Circuits have non-linearities
  - hard: e.g. supply clipping
  - weak:  $y_{out} = G_1 \cdot x_{in} + G_2 \cdot x_{in}^2 + G_3 \cdot x_{in}^3 + \dots$   
 $G_1 \gg G_2 \ \& \ G_1 \gg G_3$
- Effects:
  - Gain compression
  - Blocking & Desensitization
  - Inter-modulation: IP2 & IP3
- Cascade of stages

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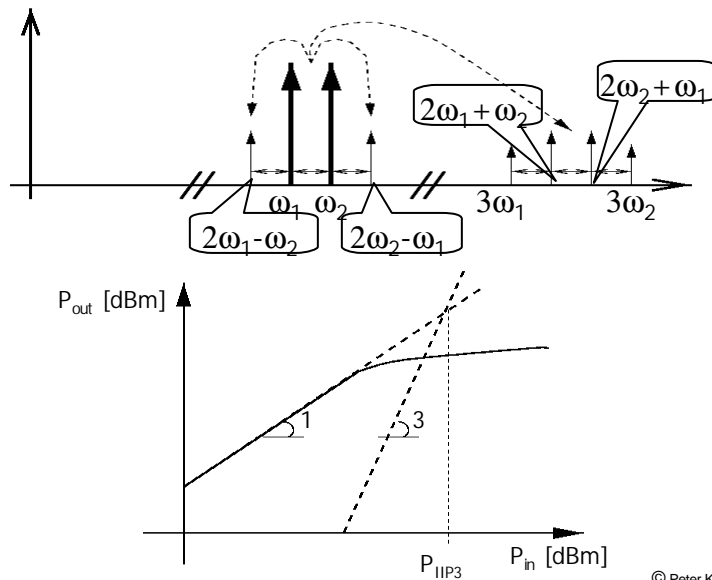
## Gain Compression



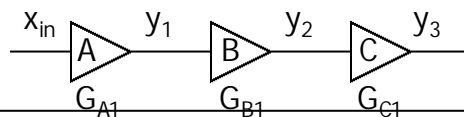
## Inter-modulation: 2nd order



## Inter-modulation: 3rd order



## IIP<sub>3</sub> for a cascade of stages



$$\frac{1}{A_{IIP3}^2} \leq \frac{1}{A_{AIP3}^2} + \frac{G_{A1}^2}{A_{BIP3}^2} + \frac{G_{A1}^2 \cdot G_{B1}^2}{A_{CIP3}^2}$$

- worst-case approximation for narrow band systems !
- voltage/current levels and gains
- effect of non-linearities more important at *later stages* !

## Spurious Free Dynamic Range

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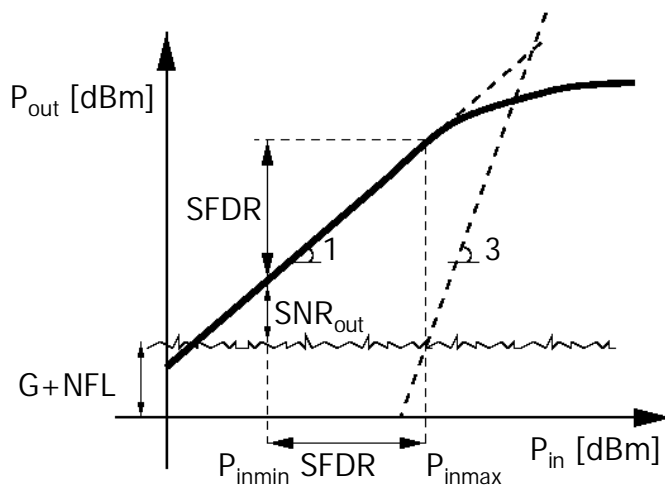
$$\text{dynamic range} = \frac{\text{max. input level}}{\text{min. input level}}$$

- under certain conditions:
  - min. level such that  $(S/N)_{\text{out}}$  is sufficient
  - max. level such that:
    - effects of non-linearities are  $\leq$  noise
    - i.e. IM3 products  $\leq$  noise
- other applications use different conditions

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## Spurious Free Dynamic Range

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# Wireless Communication Systems

## **Wireless Communications - Overview**

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- 'ether' is one medium shared by all
- 1<sup>st</sup> problem: Duplexing
  - how to arrange for a two way communication link
- 2<sup>nd</sup> problem: Multiple Access
  - how to arrange for multiple users

## Duplexing - Overview

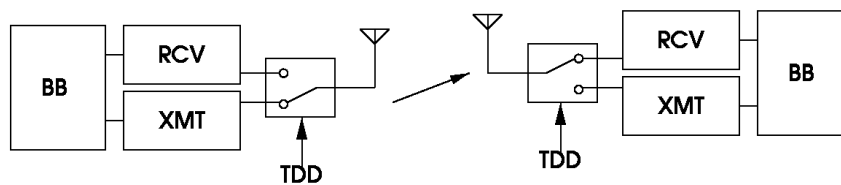
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- Establish two way communications:
  - Time division duplex:
    - same rcv and xmt frequency channel
    - alternating in time between rcv & xmt
  - Frequency division duplex
    - different frequency channel for rcv and xmt
    - full duplex possible

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## Time Division Duplex (TDD)

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- peer to peer communications
- antenna switch

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## TDD design issues

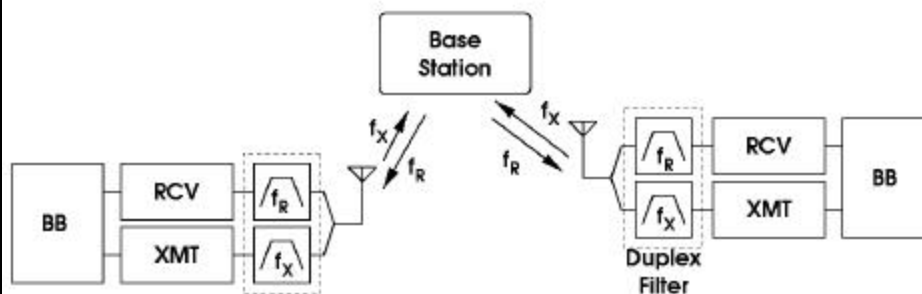
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- + mobile units can communicate
- + Switch low loss ( $< 1\text{dB}$ )
- + XMT cannot desensitize RCV
- nearby XMT can overload RCV
- + channel leakage from P/A reduction by proper timing
- packet based communication:
  - Synchronization & Buffering needed
  - digital implementation

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## Frequency Division Duplex (FDD)

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- base station  $\leftrightarrow$  mobile unit
- no peer to peer communication
- duplex filter

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## FDD design issues

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- duplexer loss (2~3dB)
  - adds directly to noise figure
  - reduces XMT efficiency
- duplexer isolation < ~50dB
  - still desensitization of RCV by XMT possible
- + less sensitive to nearby XMT
- direct XMT antenna connection
  - LO transients or P/A switch results in channel leakage
- + analog implementation

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## Multiple Access - Overview

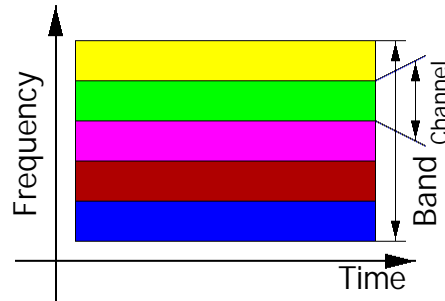
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- Frequency Division Multiple Access (FDMA)
  - divide band in channels & allocate *different channel* for each user
- Time Division Multiple Access (TDMA)
  - same channel for different users but each user accesses in a *different time-slot*
- Code Division Multiple Access (CDMA)
  - all users use same channel at same time but have a *different code*
- Carrier Sense Multiple Access (CSMA)
  - all users use same channel at *different (random) times*

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## Frequency Division MA (FDMA)

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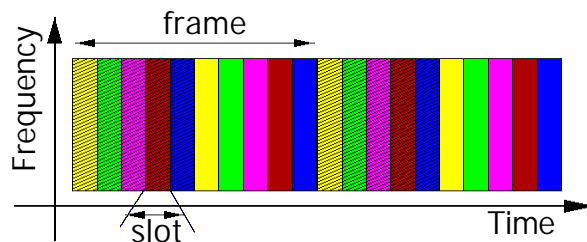


- each user is assigned a channel
- FDD & FDMA → xmt & rcv channel
- + implementation can be done analog
- you need high quality filters (loss...)

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## Time Division Multiple Access (TDMA)

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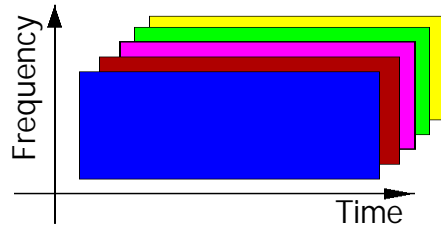


- each user is assigned a slot
- synchronization & data buffering → digital
- + add coding, correction, compression → capacity ↑
- + FDD & TDMA:
  - time RCV & XMT non-simultaneous
  - advantages of TDD

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## Code Division Multiple Access (CDMA)

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- each user has different code
  - ~ speaks different language
- Direct Sequence Spread Spectrum
  - code used to encode data
- Frequency Hopping Spread Spectrum
  - code used to select frequency sequence

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## Carrier sense multiple access (CSMA)

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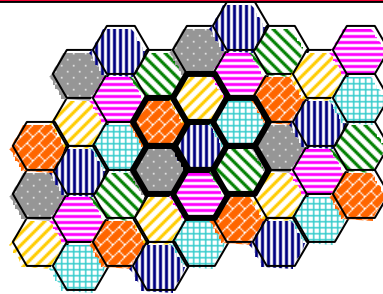
- sense medium before transmit
  - if free, transmit information
  - if collision, back-off and re-send information
- system implications similar to TDMA
- BUT,
  - + no synchronization necessary
  - no guaranteed bandwidth
    - ➔ used for data communications
    - e.g. wireless LAN

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## Cellular Communications System

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- large number of users
- cellular system
  - stations far enough
    - ➔ frequency reuse
  - far ~ transmitted power
- Co-channel interference
  - ~ distance 2 co-channel cells/cell radius
  - power independent
  - 7 reuse: ratio = 4.6 (18dB)
- Base-station & mobile unit
  - forward/up link: base ➔ mobile
  - reverse/down link: mobile ➔ base
  - hand-off: switch base stations



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## Channel characteristics

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- Path-loss:
  - propagation characteristics
- Multi-path fading:
  - direct & reflected signals interfere at rcv
- Delay Spread:
  - direct & delayed signals interfere
- ➔ fast & large variations in signal strength in moving receiver
- ➔ “frequency blocking” in stationary receiver

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## Standards - Some Examples

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- Advanced Mobile Phone Service (AMPS)
- North American Digital Standard (NADS) IS-54
- IS-95 DS CDMA - Qualcomm CDMA
- Global System for Mobile Communications (GSM)
- Digital Enhanced Cordless Telephone (DECT)
- IEEE 802.11
- HiperLAN
- .....

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## GSM

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- **G**lobal **S**ystem for **M**obile Communications
- FDD:
  - RCV: 935-960 MHz
  - XMT: 890-915 MHz
- FDMA & TDMA:
  - 200 kHz Channels
  - frame = 8 slots: 4 rcv & 4 xmt
  - RCV & XMT slot offset by 3 time slots
  - data rate ~ 270kbits/sec
- GMSK modulation
  - constant envelope - BT=0.3

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## GSM Type approval (summary)

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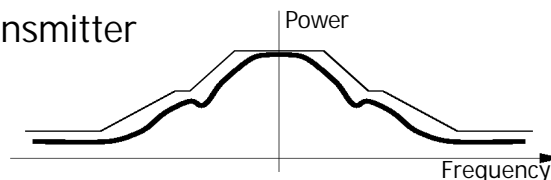
- Receiver
  - BER  $\sim 10^{-3}$  or S/N @ demodulator > 9dB
  - signal range: -102dBm to -15dBm
  - for signal of -99dBm:
    - blocking: in band: -43 up to -23dBm  
out of band: 0dBm
    - inter-modulation: -49dBm @800kHz & @1600kHz
  - for signal of -82dBm:
    - co-channel test: 9dB smaller interferer in same channel
    - adjacent channel (@200kHz): 9dB larger
    - alternate channel (@400kHz): 41dB larger

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## GSM Type approval (summary)

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- Transmitter



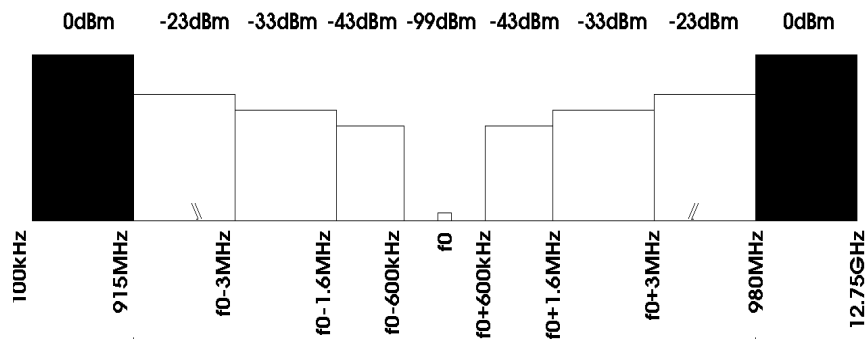
- close-in: modulation spectrum (spectral mask)
- wide-band: noise spectrum e.g.
  - noise@3MHz < -115dBc/Hz
  - noise@6MHz < -130dBc/Hz
  - noise@25MHz < -130/-136dBc/Hz
- average phase error < 5 deg.RMS
- output power
  - up to 2-3 Watt: 33-35dBm
  - power control: 28dB
- carrier leakage < 40dBc

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## Receivers

### Radio Receiver Problem (e.g. GSM)

- small signal: down to -102dBm
- narrow band signal: 200kHz on ~900MHz
- very hostile environment ➔ interference
  - e.g. blocking signals ~**100dB** larger than signal !!



## Filter as RCV

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- e.g. GSM  $f_0=900\text{MHz}$   
 $\text{BW}=200\text{kHz}$
- Quality factor:  $\sim 4500$ 
  - high Q  $\Rightarrow$  high loss  $\Rightarrow$  high NF
- High rejection & sharp filter
- Tunable filter
  - center frequency accuracy

**No Filter Technology available**

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## Heterodyne Receiver

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- down-convert signal to lower fixed intermediate frequency (IF) for filtering
  - $\Rightarrow$  Q lower
  - $\Rightarrow$  fixed frequency

- Mixer

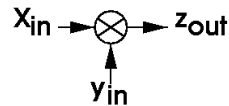
- $Z_{\text{out}} = K \cdot X_{\text{in}} \cdot Y_{\text{in}}$

- frequency translation:

- $X_{\text{in}}@ \omega_1$  &  $Y_{\text{in}}@ \omega_2 \Rightarrow Z_{\text{out}}@ |\omega_2 \pm \omega_1|$

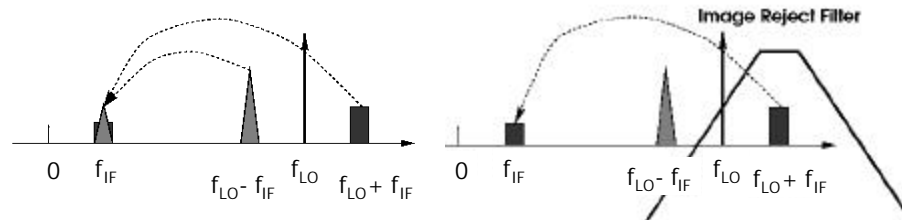
- conversion gain:

- $\text{CVG} = Z_{\text{out}} / X_{\text{in}} = K \cdot Y_{\text{in}}$



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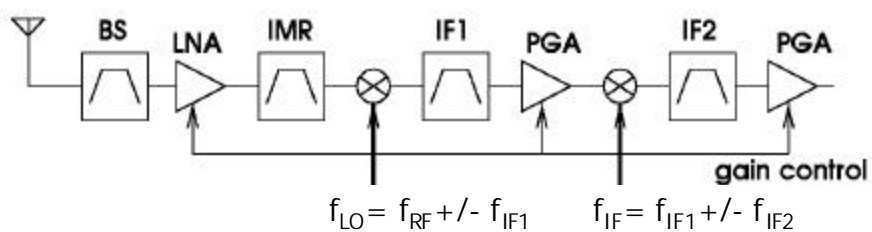
## Heterodyne Receiver: IMAGES ....



- $f_o + f_{IF}$  &  $f_o - f_{IF}$  mix with  $f_o$  to same  $f_{IF}$
- potential interference
- add IMAGE REJECT FILTER before mixer

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## Heterodyne: choice of IF



- high IF + more relaxed image filter  
+ smaller IF filter  
- higher Q  $\rightarrow$  higher loss
- multiple IFs: distribute channel filtering
- filter-amplify-filter-amplify
- gain at different frequencies: no oscillation risk

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## Mixer Spurious Responses

- image frequency
- feed-through to IF: (LO  $\rightarrow$  IF and RF  $\rightarrow$  IF)
- mixer: never only second but also higher order
  - e.g. spurious response table for double balanced mixer

		$f_{LO}$	$2f_{LO}$	$3f_{LO}$	$4f_{LO}$	$5f_{LO}$	$6f_{LO}$
6	$f_{RF}$	-100	-92	-97	-95	-100	-100
5	$f_{RF}$	-90	-84	-86	-72	-92	-70
4	$f_{RF}$	-90	-84	-97	-86	-97	-90
3	$f_{RF}$	-75	-63	-66	-72	-72	-58
2	$f_{RF}$	-70	-72	-72	-70	-82	-62
1	$f_{RF}$	-60	0	-35	-15	-37	-37
			-60	-60	-70	-72	-72

- frequency planning

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## Frequency Planning: spurious responses

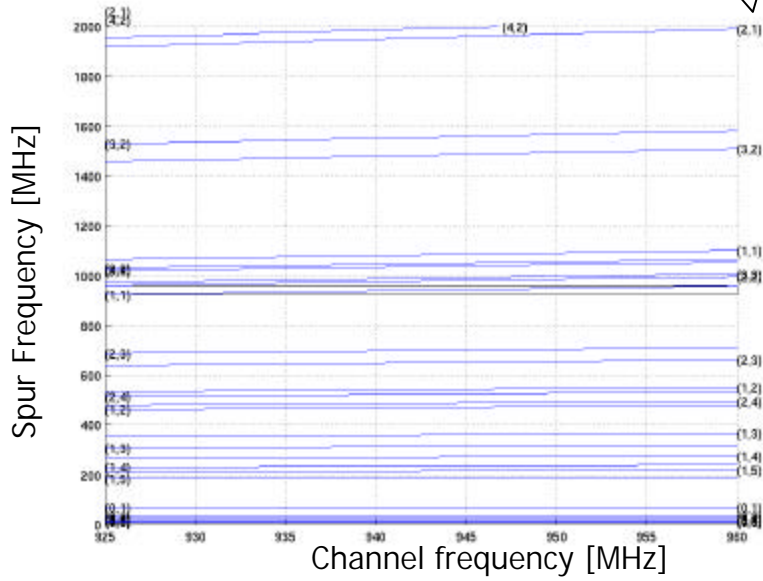


- e.g. low side injection difference mixer
  - $f_{IF} = f_{LO} - f_{RF}$
  - e.g. GSM RCV
    - RF in: 925-960MHz
    - IF: 71MHz
    - LO: 996-1031MHz
- find all spur frequencies  $f_s$ 
  - $|n f_s \pm m f_{LO}| = f_{IF}$
  - $n: 0, 1, 2, \dots; m: 0, 1, 2, \dots$

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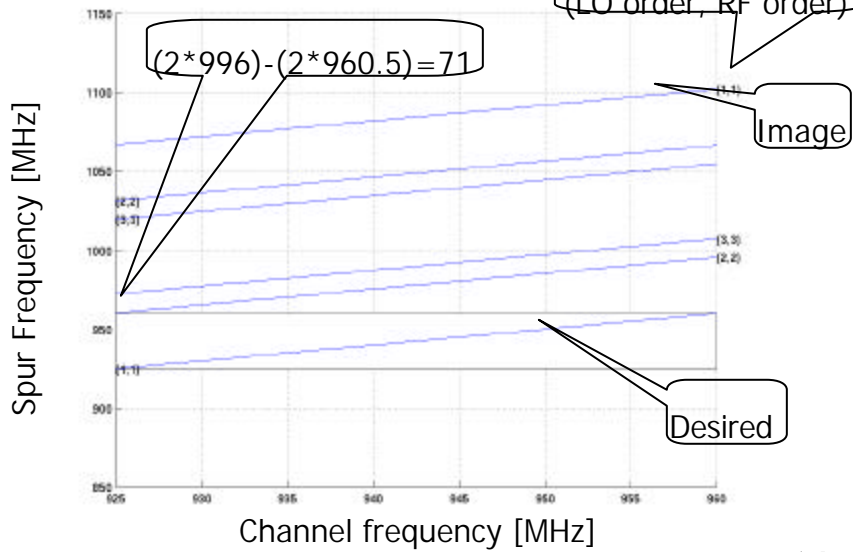
## Spurious Responses

(LO order, RF order)

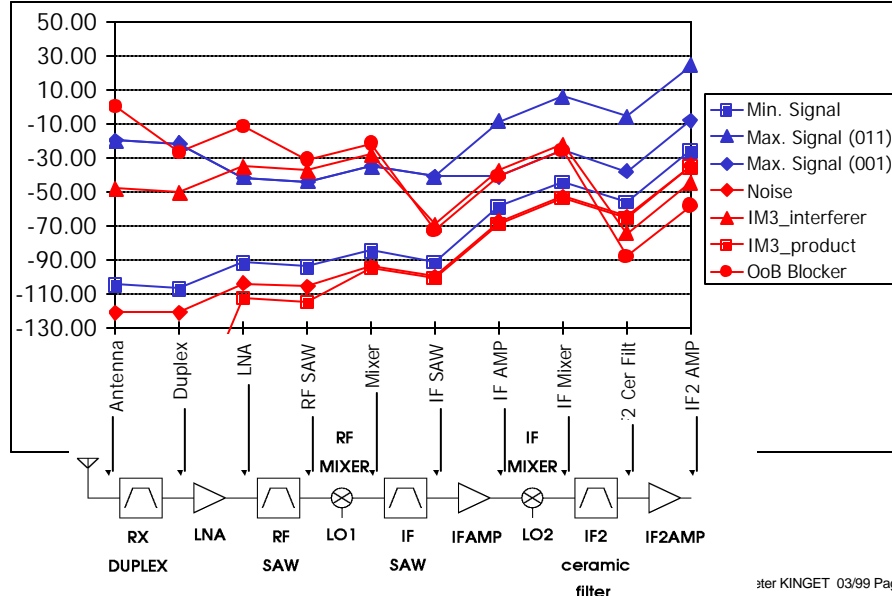


## Spurious Responses (zoom)

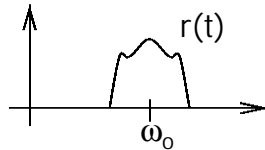
(LO order, RF order)



## Level Diagram



## Band-limited signal: Complex envelope

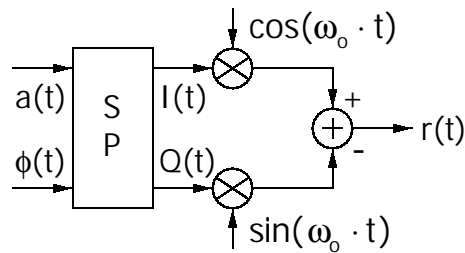


$$r(t) = a(t) \cdot \cos(\omega_0 \cdot t + \phi(t))$$

$$r(t) = I(t) \cdot \cos(\omega_0 \cdot t) - Q(t) \cdot \sin(\omega_0 \cdot t)$$

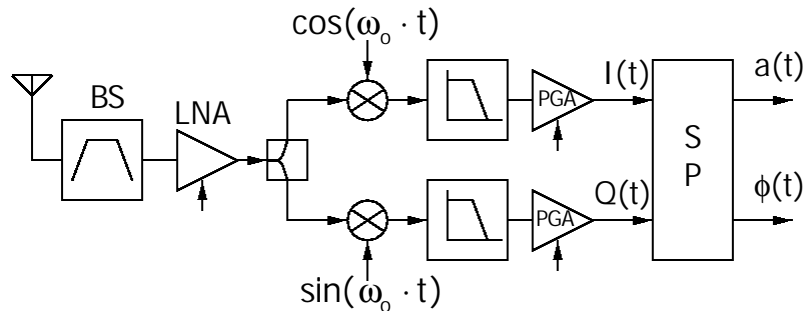
$$a(t) = \sqrt{I(t)^2 + Q(t)^2}$$

$$\phi(t) = \tan^{-1}\left(\frac{Q(t)}{I(t)}\right)$$



## Homodyne Receiver

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- $f_{LO} = f_{RF} \Rightarrow f_{IF} = 0$
- image = signal
- quadrature down-converter
- lowpass filter does channel selection

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## Homodyne design issues (1)

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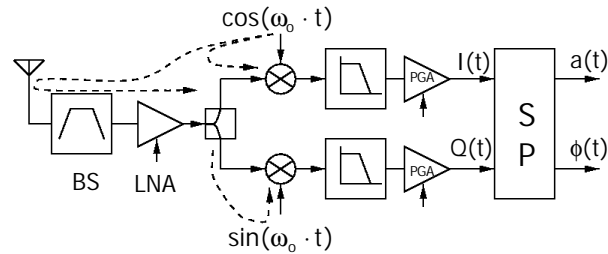
- Lowpass filters for channel selection
  - can be integrated on IC
  - high dynamic range required
    - preceded by limited gain or filtering
  - a lot of (programmable) gain at DC
    - parasitic feedback can cause stability problems
  - DC offset
  - 1/f noise

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## Homodyne design issues (2)

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- Time-varying DC offsets
  - self-mixing
    - LO leakage
    - RF leakage
- LO emission
- I/Q mismatches

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## Homodyne design issues (3)

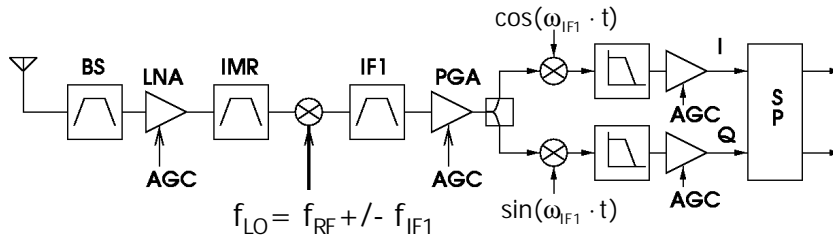
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- Even order distortion
  - IM2@LNA -> LF signal -> mixer RF/IF feed-through
  - IM2@Mixer -> LF signal & DC
  - differential circuits
  - but P/A single-ended -> antenna SE -> LNA SE
  - single-ended to differential conversion at RF ....

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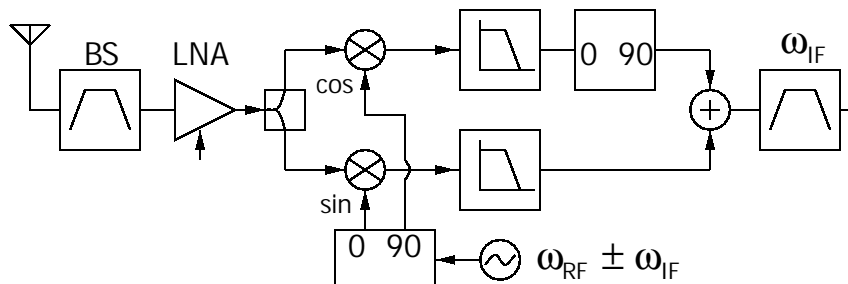
## Why not for IF

- Passive IF filters: high DR
- DC offset out of band: ac coupling
- IM2 out of band: ac coupling
- @IF
  - 1/f noise low
  - DC offset out of band
- $f_{LO} = f_{RF} \pm f_{IF}$  : emission filtered
- Modern IF: zero-IF back-end to go into DSP



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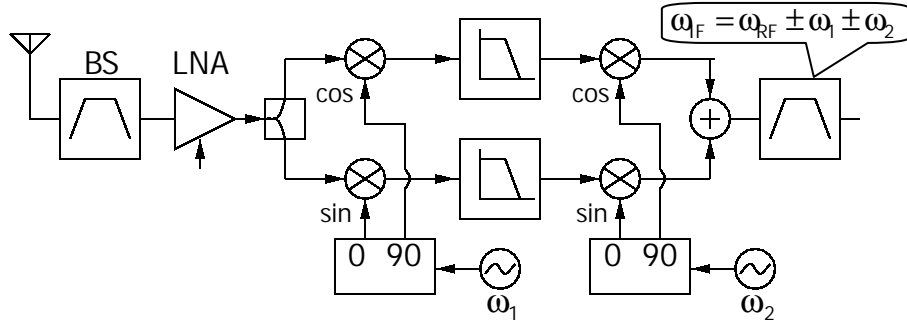
## Image Reject Receiver: Hartley



- no IMR filter
- image rejection depends on
  - quadrature accuracy
  - gain matching
- 90 degrees shift in signal path

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## Image Reject Receiver: Weaver



- use 2<sup>nd</sup> quadrature mixing stage instead of 90deg. shift
- additional secondary image

## Transmitters

## Transmitters - Overview

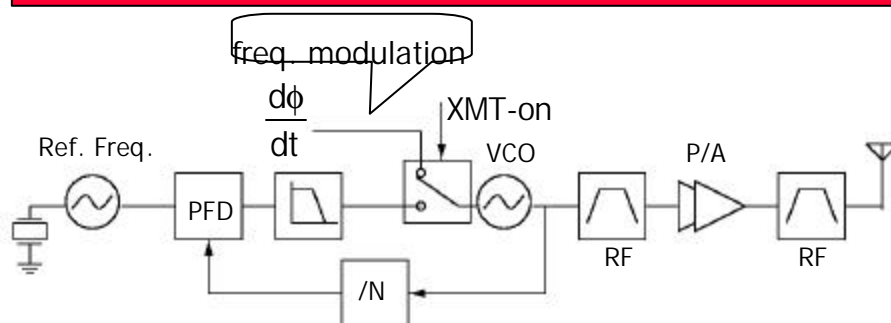
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- Basic functions:
  - modulation:
    - encode the information on a waveform's amplitude, phase or frequency
  - up-conversion:
    - move signal to desired RF carrier frequency
  - power amplification
    - amplify signal to deliver wanted power to antenna for emission

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## Direct VCO modulation

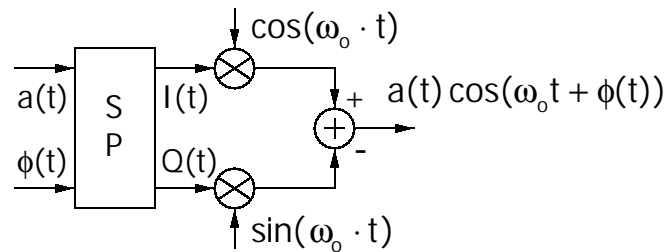
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- only constant envelope modulation
- VCO in open loop during XMT
  - frequency drift
  - pushing/pulling
  - close-in VCO noise
  - switch time XMT/RCV includes lock time
- compact

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## Quadrature Modulator

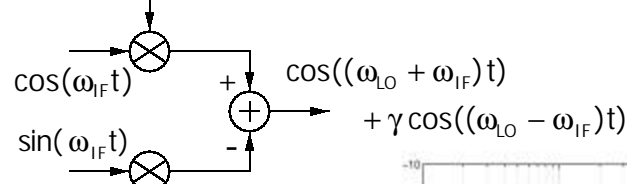


- Any modulation format
  - see complex envelope
- But unwanted sideband when
  - non perfect quadrature
  - gain mismatches

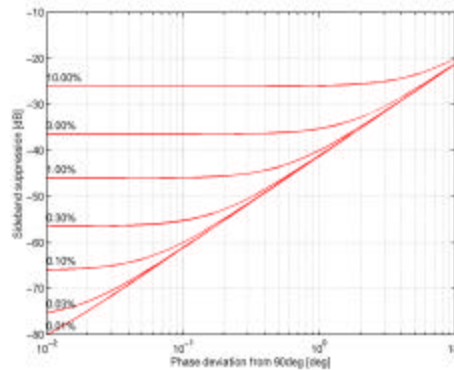
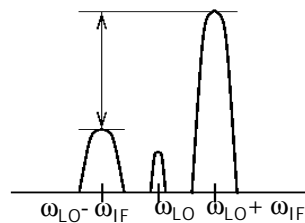
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## Quadrature modulator: Side-band rejection

$$(1 + \Delta/2) \cos(\omega_{LO} t + \Delta\phi/2)$$

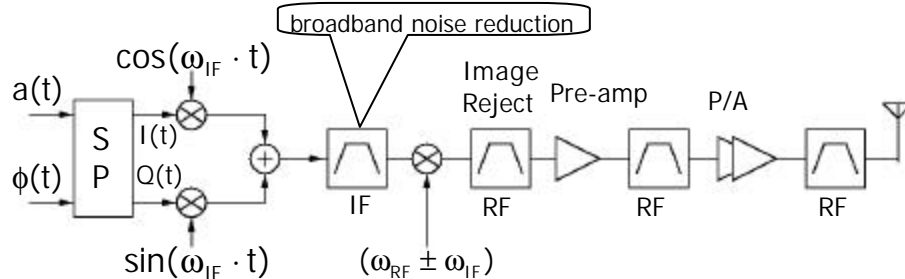


$$(1 - \Delta/2) \sin(\omega_{LO} t - \Delta\phi/2)$$



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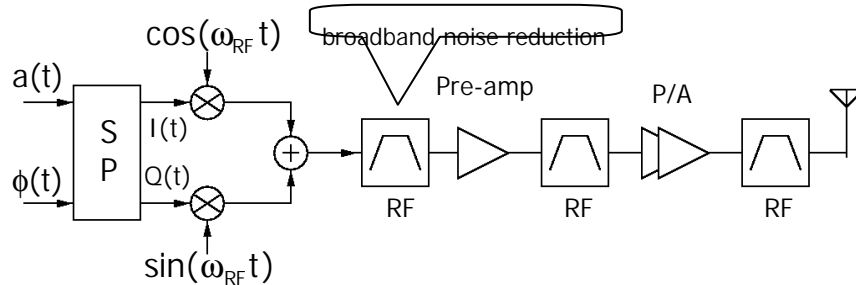
## Multi-step Up-conversion



- good image reject filter necessary
- potential for other spurs
- extra filter to reject broadband noise

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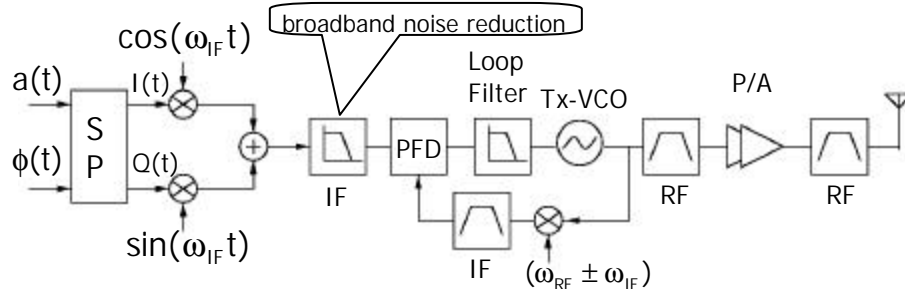
## Direct Up-conversion



- no IF and no spurs: relaxed filtering
- extra filter to reject broadband noise
- potential RF VCO re-modulation by P/A out
  - VCO shielding
- quadrature RF signal required

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## Indirect VCO modulation



- only constant envelope modulation
- loop filter BW > signal BW
- low broadband noise !
- Tx-VCO: high power & low noise (e.g.  $P_{out}$  10dBm typ. in GSM)
- potential for spurs

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## Power amplifier & output filters

- TDD: P/A - switch - antenna
  - ~1dB loss in switch
- FDD: P/A - duplexer - antenna
  - ~2-3dB loss in switch
  - 30-50% of P/A power dissipated in duplexer
- average efficiency P/A << 50%
  - depends strongly on modulation format
- $P_{out}/P_{DC}$  << 25%

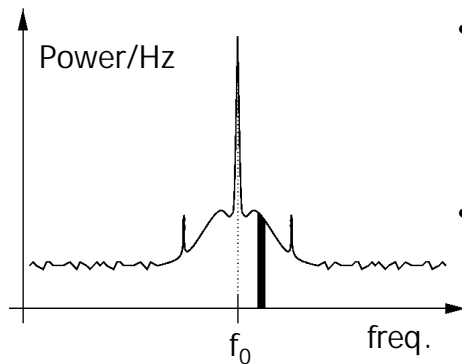
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## Transceiver design

### Frequency Synthesizer

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- 3<sup>rd</sup> subsystem in transceiver



- RCV:
  - phase noise level in side bands
  - discrete spurs
- XMT:
  - RMS phase error = integrated phase noise
  - wideband noise
  - discrete spurs



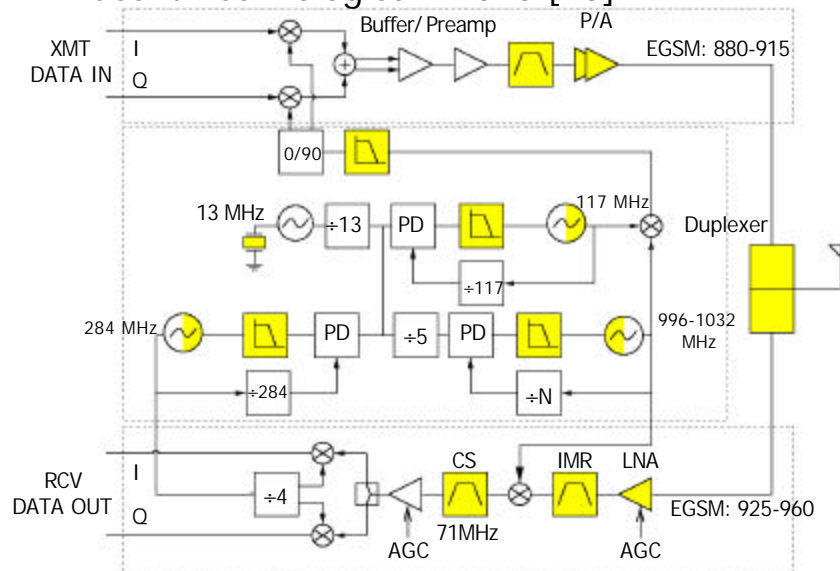
## Transceiver Design

- **Meet the standard !!!!**
- Architecture selection and system design
  - Bill of materials
  - Frequency planning:
    - # VCOs & spurious responses
  - Power consumption:
    - Transmitter (P/A) talk time
    - Receiver standby time
  - Partitioning
    - Hardware/Software
    - Analog/Digital
- Time to market & Price & Package

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## GSM Transceiver Example

- Lucent Technologies W2020 [10]



## Recent Transceiver Architectures

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- Some Trends:
  - integration & cost reduction
  - dual band
  - multi standard
- Some Techniques
  - Zero-IF
  - Low-IF
  - Double Low-IF
  - Wide-band-IF
  - IF sampling
  - $\Delta$ - $\Sigma$  decimation filter as channel select
  - Software Radio
  - .....

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## Acknowledgments

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- I would like to thank the following colleagues for stimulating discussions:
  - Kirk Ashby, Mihai Banu, Paul Davis, Jack Glas, Venu Gopinathan

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