EE E6820: Speech & Audio Processing & Recognition

# Lecture 5: Speech modeling

- 1 Modeling speech signals
- 2 Spectral and cepstral models
- 3 Linear Predictive models (LPC)
- 4 Other signal models
- 5 Speech synthesis

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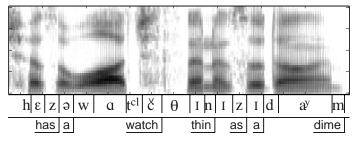
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# The speech signal

Speech sounds in the spectrogram



- Elements of the speech signal:
  - spectral resonances (formants, moving)
  - periodic excitation (voicing, pitched)
    - + pitch contour
  - noise excitation (fricatives, unvoiced, no pitch)
  - transients (stop-release bursts)
  - amplitude modulation (nasals, approximants)
  - timing!

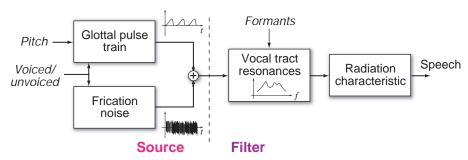


#### The source-filter model

Notional separation of:

**source:** excitation, fine time-frequency structure

& filter: resonance, broad spectral structure



• More a modeling approach than a single model

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# Signal modeling

- Signal models are a kind of representation
  - to make some aspect explicit
  - for efficiency
  - for flexibility
- Nature of model depends on goal
  - classification: remove irrelevant details
  - coding/transmission: remove perceptual irrelevance
  - modification: isolate control parameters
- But commonalities emerge
  - perceptually irrelevant detail (coding)
     will also be irrelevant for classification
  - modification domain will usually reflect 'independent' perceptual attributes
  - getting at the abstract information in the signal

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### Different influences for signal models

#### Receiver:

- see how signal is treated by listeners
  - → cochlea-style filterbank models ...

#### • Transmitter (source)

- physical vocal apparatus can generate only a limited range of signals...
  - → LPC models of vocal tract resonances

#### Making explicit particular aspects

- compact, separable correlates of resonances
  - → cepstrum
- modeling prominent features of NB spectrogram
  - → sinusoid models
- addressing unnaturalness in synthesis
  - → Harmonic+Noise model

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### Applications of (speech) signal models

- Classification / matching
  - Goal: highlight important information
  - speech recognition (lexical content)
  - speaker recognition (identity or class)
  - other signal classification
  - content-based retrieval
- Coding / transmission / storage

#### Goal: represent just enough information

- real-time transmission e.g. mobile phones
- archive storage e.g. voicemail
- Modification/synthesis

#### Goal: change certain parts independently

- speech synthesis / text-to-speech (change the words)
- speech transformation / disguise (change the speaker)



#### **Outline**

- 1 Modeling speech signals
- Spectral and cepstral models
  - Auditorily-inspired spectra
  - The cepstrum
  - Feature correlation
- 3 Linear predictive models (LPC)
- 4 Other models
- 5 Speech synthesis

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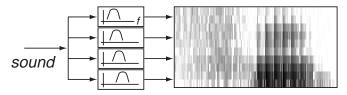
# Spectral and cepstral models

- Spectrogram seems like a good representation
  - long history
  - satisfying in use
  - experts can 'read' the speech
- What is the information?
  - intensity in time-frequency cells;
     typically 5ms x 200 Hz x 50 dB
- → Discarded detail:
  - phase
  - fine-scale timing
- The starting point for other representations



# The filterbank interpretation of the short-time Fourier transform (STFT)

 View spectrogram rows as coming from separate bandpass filters:



· Mathematically:

$$\begin{split} X[k,n_0] &= \sum_n x[n] \cdot w[n-n_0] \cdot \exp{-j\left(\frac{2\pi k(n-n_0)}{N}\right)} \\ &= \sum_n x[n] \cdot h_k[n_0-n] \\ \text{where } h_k[n] &= w[-n] \cdot \exp{j\left(\frac{2\pi kn}{N}\right)} \end{split}$$

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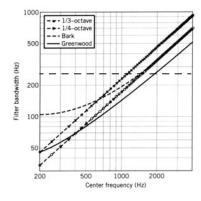
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# Spectral models: Which bandpass filters?

- Constant bandwidth? (analog / FFT)
- But: cochlea physiology & critical bandwidths
  - → implement ear models with bandpass filters & choose bandwidths by e.g. CB estimates
- Auditory frequency scales
  - constant 'Q' (center freq/bandwidth), mel, Bark...



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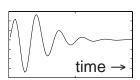
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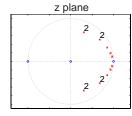
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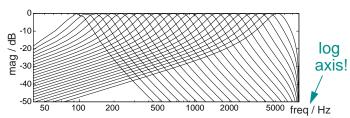
#### **Gammatone filterbank**

- · Given bandwidths, which filter shapes?
  - match inferred temporal integration window
  - match inferred spectral shape (sharp hi-F slope)
  - keep it simple (since it's only approximate)
- → Gammatone filters

$$h[n] = n^{N-1} \cdot \exp-bn \cdot \cos(\omega_i n)$$







- 2N poles, 2 zeros, low complexity
- reasonable linear match to cochlea

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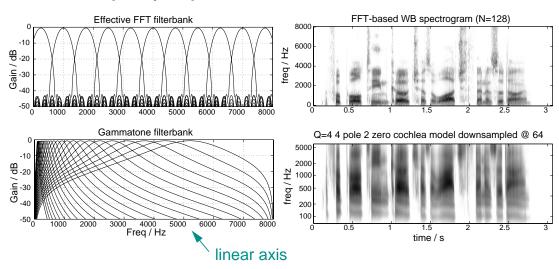
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#### Constant-BW vs. cochlea model

• Frequency responses:

#### • Spectrograms:



Magnitude smoothed over 5-20 ms time window

# **Limitations of spectral models**

- Not much data thrown away
  - just fine phase/time structure (smoothing)
  - little actual 'modeling'
  - still a large representation!
- Little separation of features
  - e.g. formants and pitch
- Highly correlated features
  - modifications affect multiple parameters
- But, quite easy to reconstruct
  - iterative reconstruction of lost phase

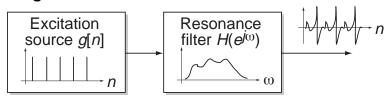
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# The cepstrum

• Original motivation: Assume a source-filter model:

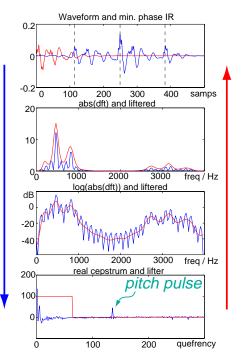


- Define 'Homomorphic deconvolution':
  - source-filter convolution: g[n]\*h[n]
  - FT  $\rightarrow$  product  $G(e^{j\omega}) \cdot H(e^{j\omega})$
  - $\log \rightarrow \text{sum}$ :  $\log G(e^{j\omega}) + \log H(e^{j\omega})$
  - IFT
    - $\rightarrow$  separate fine structure:  $c_g[n] + c_h[n]$
    - = deconvolution
- Definition:

 $\text{Real cepstrum } c_n = \operatorname{idft}(\log|\operatorname{dft}(x[n])|)$ 

### Stages in cepstral deconvolution

- Original waveform has excitation fine structure convolved with resonances
- DFT shows harmonics modulated by resonances
- Log DFT is sum of harmonic 'comb' and resonant bumps
- IDFT separates out resonant bumps (low quefrency) and regular, fine structure ('pitch pulse')
- Selecting low-n cepstrum separates resonance information (deconvolution / 'liftering')



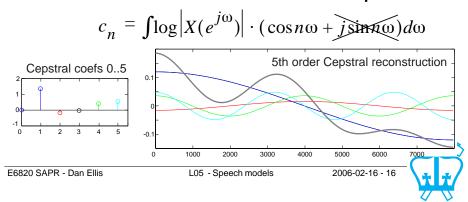
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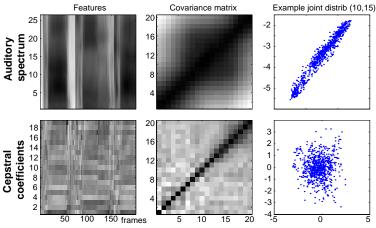
# Properties of the cepstrum

- Separate source (fine) & filter (broad structure)
  - smooth the log mag. spectrum to get resonances
- Smoothing spectrum is filtering along freq.
  - i.e. convolution applied in Fourier domain
     → multiplication in IFT ('liftering')
- Periodicity in time → harmonics in spectrum
   'pitch pulse' in high-n cepstrum
- Low-n cepstral coefficients are DCT of broad filter / resonance shape:



#### **Aside: Correlation of elements**

- Cepstrum is a popular in speech recognition
  - feature vector elements are decorrelated:



- c<sub>0</sub> 'normalizes out' average log energy
- Decorrelated pdfs fit diagonal Gaussians
  - simple correlation is a waste of parameters
- DCT is close to PCA for (mel) spectra?

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#### **Outline**

- 1 Modeling speech signals
- 2 Spectral and cepstral modes
- 3 Linear Predictive models (LPC)
  - The LPC model
  - Interpretation & application
  - Formant tracking
- 4 Other models
- 5 Speech synthesis



# 3

# **Linear predictive modeling (LPC)**

- LPC is a very successful speech model
  - it is mathematically efficient (IIR filters)
  - it is remarkably accurate for voice (fits source-filter distinction)
  - it has a satisfying physical interpretation (resonances)
- Basic math

- model output as linear function of prior outputs:

$$s[n] = (\sum_{k=1}^{p} a_k \cdot s[n-k]) + e[n]$$

... hence "linear prediction" (pth order)

- e[n] is excitation (input), a/k/a prediction error

$$\Rightarrow \frac{S(z)}{E(z)} = \frac{1}{(1 - \sum_{k=1}^{p} a_k \cdot z^{-k})} = \frac{1}{A(z)}$$

... all-pole modeling,

'autoregression' (AR) model

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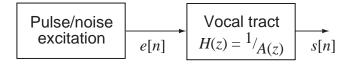
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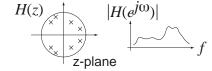


#### **Vocal tract motivation for LPC**

• Direct expression of source-filter model:

$$s[n] = (\sum_{k=1}^{p} a_k \cdot s[n-k]) + e[n]$$





- Acoustic tube models suggest all-pole model for vocal tract
- Relatively slowly-changing
  - update A(z) every 10-20 ms
- Not perfect: Nasals introduce zeros

# **Estimating LPC parameters**

Minimize short-time squared prediction error:

$$E = \sum_{n=1}^{m} e^{2}[n] = \sum_{n} \left\{ s[n] - \sum_{k=1}^{p} a_{k} s[n-k] \right\}^{2}$$

Differentiate w.r.t.  $a_k$  to get eqns for each k:

$$\sum_{n} 2(s[n] - \sum_{j=1}^{p} a_{j}s[n-j]) \cdot (-s[n-k]) = 0$$

$$\sum_{n} s[n]s[n-k] = \sum_{j} a_{j} \cdot \sum_{n} s[n-j]s[n-k]$$

$$\phi(0,k) = \sum_{j} a_{j} \cdot \phi(j,k)$$

where 
$$\phi(j,k) = \sum_{n=1}^{m} s[n-j]s[n-k]$$

are correlation coefficients

• p linear equations to solve for all  $a_i$ s...

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# **Evaluating parameters**

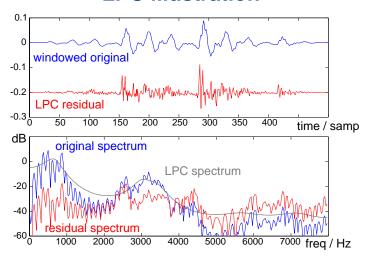
- Linear equations  $\phi(0, k) = \sum_{j=1}^{p} a_j \cdot \phi(j, k)$
- If s[n] is assumed zero outside some window  $\phi(j,k) = \sum_n s[n-j]s[n-k] = r_{ss}(|j-k|)$   $r_{ss}(\tau)$  is autocorrelation

Hence equations become:

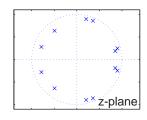
$$\begin{bmatrix} r(1) \\ r(2) \\ \dots \\ r(p) \end{bmatrix} = \begin{bmatrix} r(0) & r(1) & \dots & r(p-1) \\ r(1) & r(2) & \dots & r(p-2) \\ \dots & \dots & \dots & \dots \\ r(p-1) & r(p-2) & \dots & r(0) \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \dots \\ a_p \end{bmatrix}$$

- Toeplitz matrix (equal antidiagonals)
  - → can use Durbin recursion to solve
- (Solve full  $\phi(j, k)$  via Cholesky)

#### LPC illustration



Actual poles:



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### **Interpreting LPC**

#### • Picking out resonances

if signal really was source + all-pole resonances,
 LPC should find the resonances

#### • Least-squares fit to spectrum

- minimizing  $e^2[n]$  in time domain is the same as minimizing  $E^2(e^{j\omega})$  (by Parseval)
- →close fit to spectral *peaks*; valleys don't matter

## Removing smooth variation in spectrum

- 1/A(z) is low-order approximation to S(z)

$$-\frac{S(z)}{E(z)} = \frac{1}{A(z)}$$

- hence, residual E(z) = A(z)S(z) is 'flat' version of S

#### Signal whitening:

- white noise (independent x[n]s) has flat spectrum
- →whitening removes temporal correlation

# **Alternative LPC representations**

- Many alternate p-dimensional representations:
  - coefficients {a<sub>i</sub>}
  - roots  $\{\lambda_i\}$ :  $\prod (1 \lambda_i z^{-1}) = 1 \sum a_i z^{-1}$
  - line spectrum frequencies...
  - reflection coefficients  $\{k_i\}$  from lattice form
  - tube model log area ratios  $g_i = \log \left( \frac{1 k_i}{1 + k_i} \right)$
- Choice depends on:
  - mathematical convenience/complexity
  - quantization sensitivity
  - ease of guaranteeing stability
  - what is made explicit
  - distributions as statistics

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# **LPC Applications**

- Analysis-synthesis (coding, transmission):
  - $S(z) = \frac{E(z)}{A(z)}$

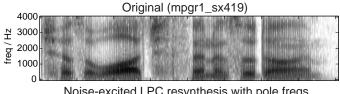
hence can reconstruct by filtering e[n] with  $\{a_i\}$ s

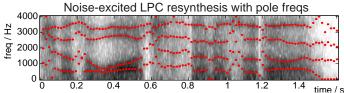
- whitened, decorrelated, minimized e[n]s are easy to quantize
- .. or can model e[n] e.g. as simple pulse train
- Recognition/classification
  - LPC fit responds to spectral peaks (formants)
  - can use for recognition (convert to cepstra?)
- Modification
  - separating source and filter supports crosssynthesis
  - pole / resonance model supports 'warping'
     (e.g. male → female)

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# **Aside: Formant tracking**

- Formants carry (most?) linguistic information
- Why not classify → speech recognition ?
  - e.g. local maxima in cepstral-liftered spectrum pole frequencies in LPC fit
- But: recognition needs to work in all circumstances
  - formants can be obscure or undefined





→ Need more graceful, robust parameters

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#### **Outline**

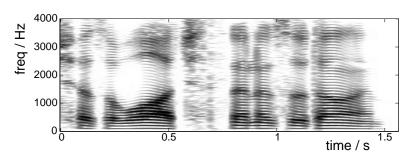
- 1 Modeling speech signals
- 2 Spectral and cepstral modes
- 3 Linear predictive models (LPC)
- Other models
  - Sinewave modeling
  - Harmonic+Noise model (HNM)
- 5 Speech synthesis



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# Other models: Sinusoid modeling

- Early signal models required low complexity
  - e.g. LPC
- Advances in hardware open new possibilities...
- NB spectrogram suggests harmonics model:



- 'important' info in 2-D surface is set of tracks?
- harmonic tracks have ~ smooth properties
- straightforward resynthesis

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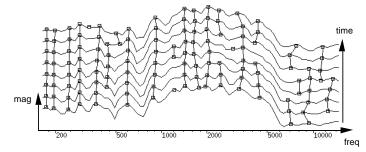


#### Sine wave models

Model sound as sum of AM/FM sinusoids:

$$s[n] = \sum_{k=1}^{N[n]} A_k[n] \cos(n \cdot \omega_k[n] + \phi_k[n])$$

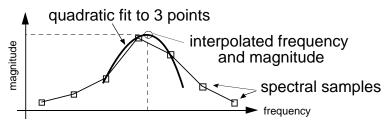
- $A_k$ ,  $\omega_k$ ,  $\phi_k$  piecewise linear or constant
- can enforce harmonicity:  $\omega_k = k.\omega_0$
- Extract parameters directly from STFT frames:



- find local maxima of |S[k,n]| along frequency
- track birth/death & correspondence

### Finding sinusoid peaks

- · Look for local maxima along DFT frame
  - i.e. |S[k-1,n]| < |S[k,n]| > |S[k+1,n]|
- Want exact frequency of implied sinusoid
  - DFT is normally quantized quite coarsely e.g. 4000 Hz / 256 bins = 15.6 Hz
  - interpolate at peaks via, e.g., quadratic fit



- may also need interpolated unwrapped phase
- Or, use differential of phase along time (pvoc):

- 
$$\omega = \frac{a\dot{b} - b\dot{a}}{a^2 + b^2}$$
 where  $S[k,n] = a + jb$ 

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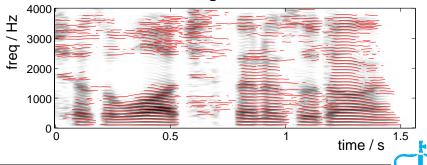
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# Sinewave modeling applications

- Modification (interpolation) & synthesis
  - connecting arbitrary  $\omega$  &  $\phi$  requires cubic phase interpolation (because  $\omega = \dot{\phi}$  )
- Types of modification
  - time & frequency scale modification
    - .. with or without changing formant envelope
  - concatenation/smoothing boundaries
  - phase realignment (for crest reduction)

• Non-harmonic signals? OK-ish



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#### Harmonics + noise model

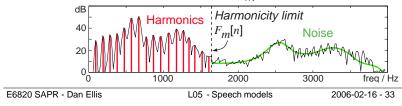
- Motivation to improve sinusoid model because:
  - problems with analysis of real (noisy) signals
  - problems with synthesis quality (esp. noise)
  - perceptual suspicions
- Model:

$$s[n] = \sum_{k=1}^{N[n]} A_k[n] \cos(n \cdot k \cdot \omega_0[n]) + e[n] \cdot (h_n[n] \otimes b[n])$$

#### **Harmonics**

**Noise** 

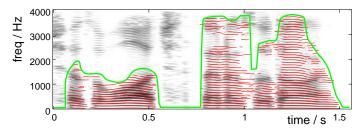
- sinusoids are forced to be harmonic
- remainder is filtered & time-shaped noise
- 'Break frequency'  $F_m[n]$  between H and N:



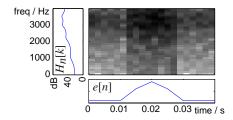
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### **HNM** analysis and synthesis

• Dynamically adjust  $F_m[n]$  based on 'harmonic test':



• Noise has envelopes in time e[n] and freq  $H_n$ 



- reconstruct bursts / synchronize to pitch pulses

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- 5 Speech synthesis
  - Phone concatenation
  - Diphone synthesis

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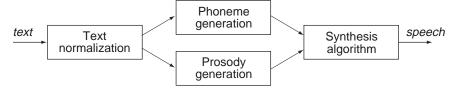
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# **Speech synthesis**

- One thing you can do with models
- Synthesis easier than recognition?
  - listeners do the work
  - .. but listeners are very critical
- Overview of synthesis

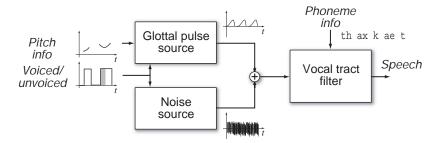


- normalization disambiguates text (abbreviations)
- phonetic realization from pronouncing dictionary
- prosodic synthesis by rule (timing, pitch contour)
- .. all controls waveform generation

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## Source-filter synthesis

Flexibility of source-filter model is ideal for speech synthesis



- Excitation source issues:
  - voiced / unvoiced / mixture ([th] etc.)
  - pitch cycle of voiced segments
  - glottal pulse shape → voice quality?

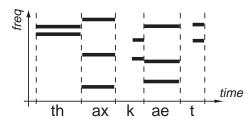
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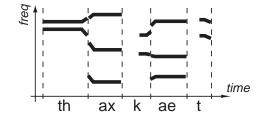


# **Vocal tract modeling**

• Simplest idea: Store a single VT model for each phoneme



- but: discontinuities are very unnatural
- Improve by smoothing between templates



- trick is finding the right domain

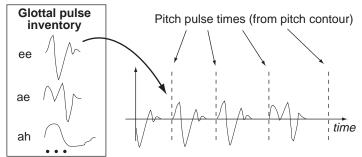
### **Cepstrum-based synthesis**

- Low-*n* cepstrum is compact model of target spectrum
- · Can invert to get actual VT IR waveform:

$$c_n = idft(log|dft(x[n])|)$$

$$\rightarrow h[n] = idft(exp(dft(c_n)))$$

- All-zero (FIR) VT response
  - → can pre-convolve with glottal pulses



cross-fading between templates is OK

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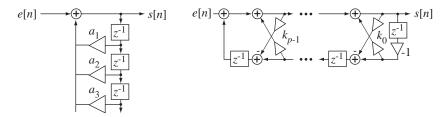
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### **LPC-based synthesis**

- Very compact representation of target spectra
  - 3 or 4 pole pairs per template
- Low-order IIR filter → very efficient synthesis
- How to interpolate?
  - cannot just interpolate  $a_i$  in a running filter
  - but: lattice filter has better-behaved interpolation



- What to use for excitation
  - residual from original analysis
  - reconstructed periodic pulse train
  - parameterized residual resynthesis

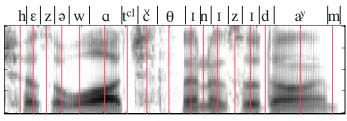
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## **Diphone synthesis**

- Problems in phone-concatenation synthesis
  - phonemes are context-dependent
  - coarticulation is complex
  - transitions are critical to perception
- → store *transitions* instead of just phonemes

#### Phones

Diphone segments



- ~40 phones → 800 diphones
- or even more context if have a larger database
- How to splice diphones together?
  - TD-PSOLA: align pitch pulses and cross-fade
  - MBROLA: normalized, multiband

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# **HNM** synthesis

- High quality resynthesis of real diphone units
  - + parametric representation for modifications
  - pitch, timing modifications
  - removal of discontinuities at boundaries
- Synthesis procedure:
  - linguistic processing gives phones, pitch, timing
  - database search gives best-matching units
  - use HNM to fine-tune pitch & timing
  - cross-fade  $A_k$  and  $\omega_0$  parameters at boundaries

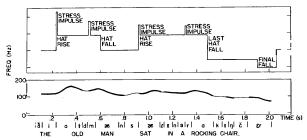


- Careful preparation of database is key
  - sine models allow phase alignment of all units
  - larger database improves unit match



## **Generating prosody**

- The real factor limiting speech synthesis?
- Waveform synthesizers have inputs for
  - intensity (stress)
  - duration (phrasing)
  - fundamental frequency (pitch)
- Curves produced by superposition of (many) inferred linguistic rules
  - phrase final lengthening, unstressed shortening..



Or learn rules from transcribed examples

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# **Summary**

- Range of models:
  - spectral, cepstral
  - LPC, Sinusoid, HNM
- Range of applications:
  - general spectral shape (filterbank) → ASR
  - precise description (LPC+residual) → coding
  - pitch, time modification (HNM) → synthesis
- Issues:
  - performance vs. computational complexity
  - generality vs. accuracy
  - representation size vs. quality

#### **Parting thought:**

not all parameters are created equal ...