

Prob I:

- a) FALSE: if $f(t)$ is odd then $f(t) = -f(-t)$ which means $-f(-t)$ is also odd
- b) FALSE: Linearity and time-invariance are two distinct concepts. For example the system of question VI is linear but not time-invariant
- c) FALSE: For most signals (the ones that satisfy Dirichlet conditions) the FS rep. is equal except at signal discontinuities
- d) TRUE: $x(t)$ is the sum of two periodic signals, with periods respectively $\sqrt{2}$ and $\frac{\sqrt{2}}{2}$. Since $\text{lcm}(\sqrt{2}, \frac{\sqrt{2}}{2}) = \sqrt{2}$ the signal $x(t)$ is periodic with period $\sqrt{2}$.
- e) FALSE: Most periodic signals have finite, but non-zero power, and therefore, since these are everlasting signals they have infinite energy.

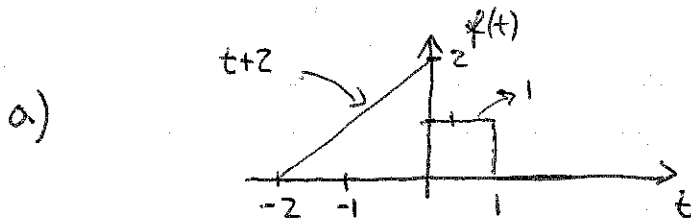
Prob II:

$$\begin{aligned}
 f(t) &= -2\cos(\omega_0 t) + 2\sqrt{3}\sin(\omega_0 t + \frac{\pi}{3}) \\
 &= -2\cos(\omega_0 t) + 2\sqrt{3}\sin(\omega_0 t)\cos(\frac{\pi}{3}) + 2\sqrt{3}\cos(\omega_0 t)\sin(\frac{\pi}{3}) \\
 &= -2\cos(\omega_0 t) + 2\sqrt{3}\sin(\omega_0 t)\frac{1}{2} + 2\sqrt{3}\cos(\omega_0 t)\frac{\sqrt{3}}{2} \\
 &= -2\cos(\omega_0 t) + \sqrt{3}\sin(\omega_0 t) + 3\cos(\omega_0 t) \\
 &= \cos(\omega_0 t) + \sqrt{3}\sin(\omega_0 t) \\
 &= a\cos(\omega_0 t) + b\sin(\omega_0 t) \quad , \text{where } a=1, b=\sqrt{3} \\
 &= C\cos(\omega_0 t + \theta) \quad , \text{where } C = \sqrt{a^2 + b^2} = \sqrt{1+3} = 2 \\
 & \quad \theta = \tan^{-1}(-\frac{b}{a}) = -\tan^{-1}(\sqrt{3}) = -\frac{\pi}{3} \\
 &= 2\cos(\omega_0 t - \frac{\pi}{3})
 \end{aligned}$$

Prob III: a) $\delta(t-3) \frac{4-jt^2}{2t} = \delta(t-3) \frac{4-j3^2}{2 \cdot 3} = \delta(t-3) \frac{4-j9}{6} = \left(\frac{2}{3} - j\frac{3}{2}\right) \delta(t-3)$

b) $\int_{-\infty}^{\infty} \sqrt{t} \cos(3\pi t) \delta(1-t) dt = \int_{-\infty}^{\infty} \sqrt{t} \cos(3\pi t) \delta(t-1) dt$
 (sampling property) $= \sqrt{1} \cos(3\pi) = \cos(+\pi) = -1$

Prob IV:

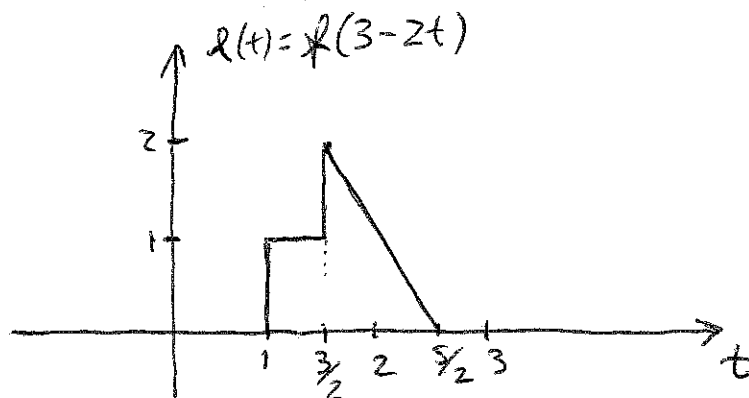
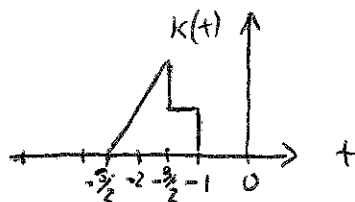
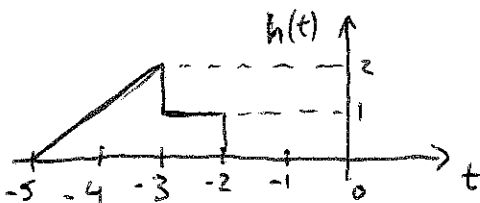


so $f(t) = (t+2)(u(t+2) - u(t)) + 1 \cdot (u(t) - u(t-1))$
 $= (t+2)u(t+2) - (t+1)u(t) - u(t-1)$

b) Let's do it in steps:

(time-shift)
 (time-scaling)
 (time-reversal)

$h(t) = f(t+3)$
 $k(t) = h(2t) = f(2t+3) = f(2t+3)$
 $l(t) = k(-t) = f(2(-t)+3) = f(-2t+3)$



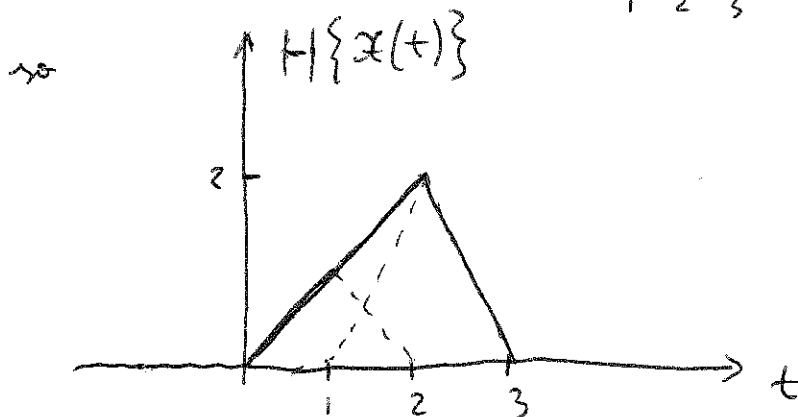
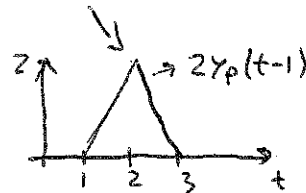
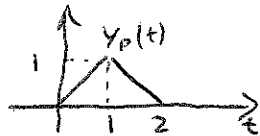
Prob V: a) $x(t)$ is clearly the sum of two pulses

$$\begin{aligned} x(t) &= (u(t) - u(t-1)) + 2(u(t-1) - u(t-2)) \\ &= p(t) + 2p(t-1) \end{aligned}$$

therefore $x(t) = ap(t) + bp(t-t_0)$ with $a=1, b=2, t_0=1$

b) Since the system is linear and time invariant we can use these properties to compute the response based on $y_p(t)$.

$$\begin{aligned} H\{x(t)\} &= H\{ap(t) + bp(t-t_0)\} \\ \text{(linearity)} \quad &= aH\{p(t)\} + bH\{p(t-t_0)\} \\ \text{(TI)} \quad &= \underbrace{a y_p(t)} + \underbrace{b y_p(t-t_0)} \end{aligned}$$



$$H\{x(t)\} = t(u(t) - u(t-2)) + (6-2t)(u(t-2) - u(t-3))$$

Prob VI: To check if the system is linear we need to check if for any inputs $x_1(t)$ & $x_2(t)$ and scalars a, b we have

$$H\{a x_1(t) + b x_2(t)\} = a H\{x_1(t)\} + b H\{x_2(t)\}$$

Let x_1, x_2 be two inputs, and y_1, y_2 be the corresponding ~~the~~ outputs, respectively. This means that

$$y_1(t) = H\{x_1(t)\} \quad y_1(t) - \frac{1}{2} y_1(t-1) = t x_1(t) \quad \forall t$$

$$y_2(t) = H\{x_2(t)\} \quad y_2(t) - \frac{1}{2} y_2(t-1) = t x_2(t) \quad \forall t$$

Let $x(t) = a x_1(t) + b x_2(t)$ if the system is linear then

$$H\{x(t)\} = y(t) = a y_1(t) + b y_2(t)$$

that is $y(t) - \frac{1}{2} y(t-1) = t x(t) \quad \forall t$

Let's check this:

$$y(t) - \frac{1}{2} y(t-1) = t x(t) \quad (\Rightarrow) \quad a y_1(t) + b y_2(t) - \frac{1}{2} (a y_1(t-1) + b y_2(t-1)) = t (a x_1(t) + b x_2(t))$$

$$(\Rightarrow) \quad a y_1(t) - a \frac{1}{2} y_1(t-1) + b y_2(t) - b \frac{1}{2} y_2(t-1) = a t x_1(t) + b t x_2(t)$$

$$(\Rightarrow) \quad a (y_1(t) - \frac{1}{2} y_1(t-1)) + b (y_2(t) - \frac{1}{2} y_2(t-1)) = a t x_1(t) + b t x_2(t)$$

$$(\Rightarrow) \quad a \underbrace{(y_1(t) - \frac{1}{2} y_1(t-1))}_{=0 \quad \forall t} - t x_1(t) + b \underbrace{(y_2(t) - \frac{1}{2} y_2(t-1))}_{=0 \quad \forall t} - t x_2(t) = 0$$

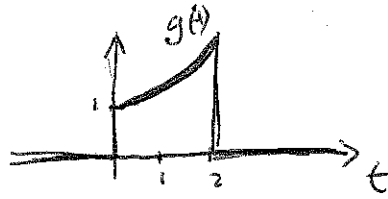
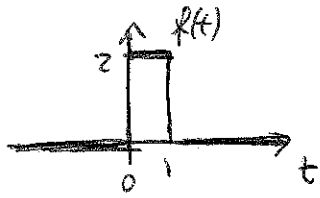
since $y_1(t) = H\{x_1(t)\}$

$y_2(t) = H\{x_2(t)\}$

So the statement is true for all t , therefore the system is linear. It is easy to check that the system is time-varying as well.

Prob VII:

a)

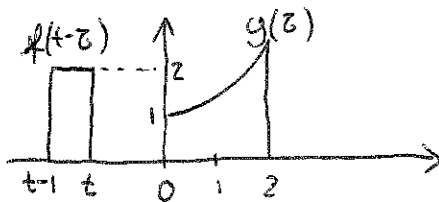


b) let's use the graphical method.

$$(f * g)(t) = \int_{-\infty}^{\infty} f(\tau) g(t-\tau) d\tau = (g * f)(t) = \int_{-\infty}^{\infty} g(\tau) f(t-\tau) d\tau$$

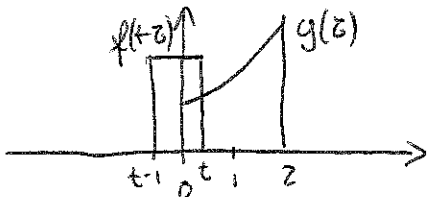
(\$f\$ is easier to "flip around")

Case 1)



~~Case 1~~ $t < 0 \Rightarrow$
 $(f * g)(t) = 0$

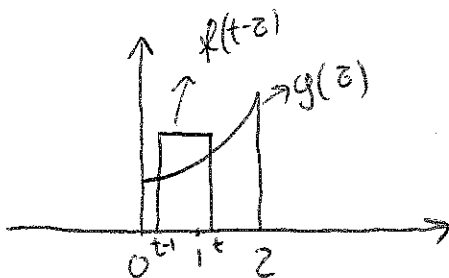
Case 2)



~~Case 2~~ $t > 0$ & $t-1 < 0 \in$
 $0 \leq t < 1$

$$(f * g)(t) = \int_0^t g(\tau) f(t-\tau) d\tau = \int_0^t e^{\tau} \cdot 2 d\tau = [2e^{\tau}]_0^t = 2e^t - 2 = 2(e^t - 1)$$

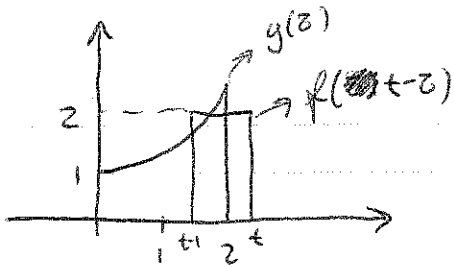
Case 3)



$t-1 \geq 0$ & $t < 2 \in$
 $1 \leq t < 2$

$$(f * g)(t) = \int_{t-1}^t e^{\tau} \cdot 2 d\tau = [2e^{\tau}]_{t-1}^t = 2e^t - 2e^{t-1} = 2e^t(1 - e^{-1})$$

Case 4)

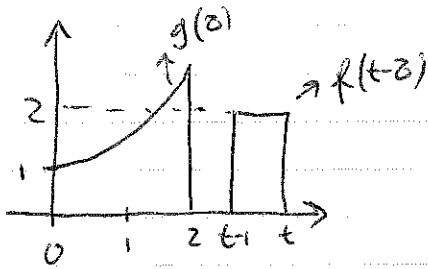


$$t \geq 2 \ \& \ t-1 < 2 \ (\Leftrightarrow)$$

$$2 \leq t < 3$$

$$(f * g)(t) = \int_{t-1}^2 2e^z dz = \left[2e^z \right]_{t-1}^2 = 2e^2 - 2e^{t-1} = 2(e^2 - e^{t-1})$$

Case 5)



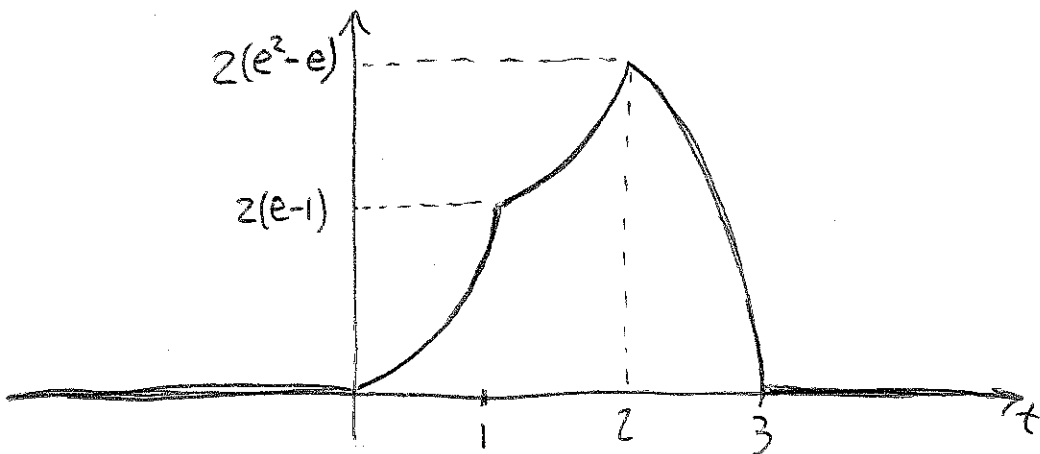
$$t-1 \geq 2 \ (\Leftrightarrow) \ t \geq 3$$

$$(f * g)(t) = 0$$

Putting all these together

$$(f * g)(t) = \begin{cases} 0 & , t < 0 \\ 2(e^t - 1) & , 0 \leq t < 1 \\ 2e^t(1 - e^{-1}) & , 1 \leq t < 2 \\ 2(e^2 - e^{t-1}) & , 2 \leq t < 3 \\ 0 & , t \geq 3 \end{cases}$$

Although it was not asked we can sketch this signal



Prob VIII:

a) Just by inspection we see that $\omega_0 = 1 \Rightarrow T_0 = 2\pi$

Another way is to notice the $x(t)$ is the sum of periodic signals with period $\frac{2\pi}{3}$ and $\frac{2\pi}{2}$, therefore it is periodic with period

$$\text{l.c.m.} \left(\frac{2\pi}{3}, \frac{2\pi}{2} \right) = 2\pi$$

period $T_0 = 2\pi$, fundamental freq. $\omega_0 = 1$

b) Note that $x(t)$ is already written in the exponential Fourier ~~series~~ form

$$x(t) = \sum_{k=-\infty}^{\infty} D_k e^{j\omega_0 k t}$$

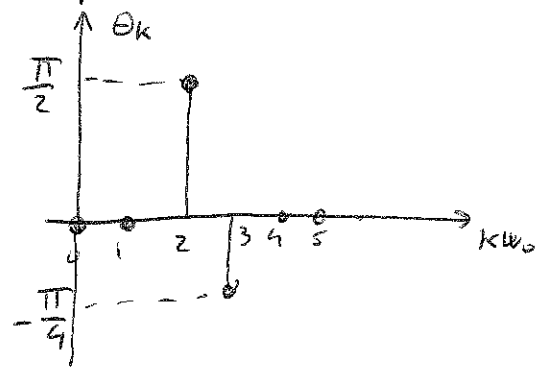
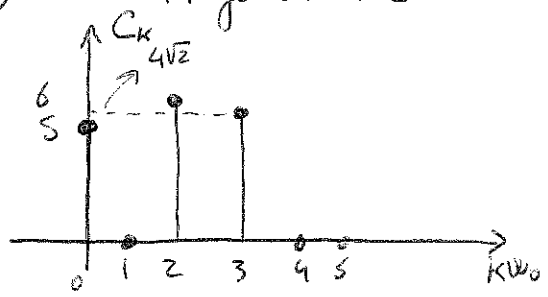
where $D_{-3} = 2+2j$, $D_{-2} = -3j$, $D_0 = 5$, $D_2 = 3j$, $D_3 = 2-2j$ and $D_k = 0$ for all other k . Now notice $D_k = D_k^*$ for all k implying the signal is real. To get the trigonometric FS representation we can either use the formulas in class, or do it directly

$$\begin{aligned}
 x(t) &= (2+2j)e^{-j3t} + (2-2j)e^{j3t} + (-3j)e^{-j2t} + 3je^{j2t} + 5 \\
 &= 2\sqrt{2}e^{j\frac{\pi}{4}}e^{-j3t} + 2\sqrt{2}e^{-j\frac{\pi}{4}}e^{j3t} + 3e^{-j\frac{\pi}{2}}e^{-j2t} + 3e^{j\frac{\pi}{2}}e^{j2t} + 5 \\
 &= 2\sqrt{2} \left(e^{j(3t-\frac{\pi}{4})} + e^{-j(3t-\frac{\pi}{4})} \right) + 3 \left(e^{j(2t+\frac{\pi}{2})} + e^{-j(2t+\frac{\pi}{2})} \right) + 5 \\
 &= 2\sqrt{2} \cdot 2\cos\left(3t - \frac{\pi}{4}\right) + 3 \cdot 2 \cdot \cos\left(2t + \frac{\pi}{2}\right) + 5 \\
 &= 4\sqrt{2} \cos\left(3t - \frac{\pi}{4}\right) + 6 \cos\left(2t + \frac{\pi}{2}\right) + 5
 \end{aligned}$$

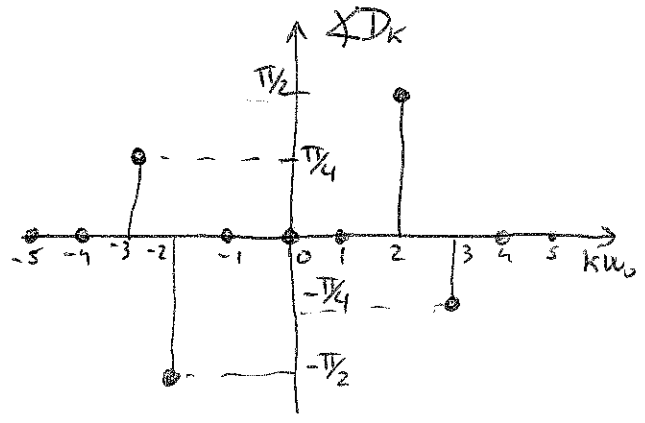
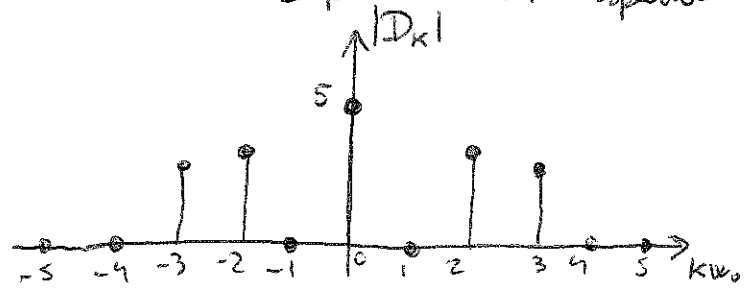
Therefore

$$x(t) = 5 + 6 \cos\left(2t + \frac{\pi}{2}\right) + 4\sqrt{2} \cos\left(3t - \frac{\pi}{4}\right)$$

c) Trigonometric Fourier spectra



Exp. Fourier spectra



d) $P_x = \sum_{k=-\infty}^{\infty} |D_k|^2$ by Parseval's theorem, therefore

$$P_x = 5^2 + 2(\sqrt{2^2+2^2})^2 + 2 \cdot 3^2 = 25 + 16 + 18 = 59$$