

3.75 Consider the two LTI causal digital filters with impulse responses given by

$$h_A[n] = 0.3\delta[n] - \delta[n-1] + 0.3\delta[n-2],$$

$$h_B[n] = 0.3\delta[n] + \delta[n-1] + 0.3\delta[n-2].$$

- (a) Sketch the magnitude responses of the two filters and compare their characteristics.  
 (b) Let  $h_A[n]$  be the impulse response of a causal digital filter with a frequency response  $H_A(e^{j\omega})$ . Define another digital filter whose impulse response  $h_C[n]$  is given by

$$h_C[n] = (-1)^n h_A[n], \quad \text{for all } n.$$

What is the relation between the frequency response  $H_C(e^{j\omega})$  of this new filter and the frequency response  $H_A(e^{j\omega})$  of the parent filter?

7.5 Let a causal LTI discrete-time system be characterized by a real impulse response  $h[n]$  with a DTFT  $H(e^{j\omega})$ . Consider the system of Figure P7.1, where  $x[n]$  is a finite-length sequence. Determine the frequency response of the overall system  $G(e^{j\omega})$  in terms of  $H(e^{j\omega})$ , and show that it has a zero-phase response.

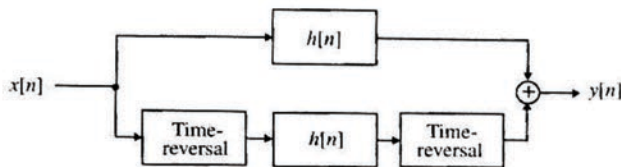


Figure P7.1

7.16 (a) Design a length-5 FIR bandpass filter with an antisymmetric impulse response  $h[n]$ , i.e.,  $h[n] = -h[4-n]$ ,  $0 \leq n \leq 4$ , satisfying the following magnitude response values:  $|H(e^{j0.3\pi})| = 0.3$  and  $|H(e^{j0.6\pi})| = 0.8$ .

(b) Determine the exact expression for the frequency response of the filter designed, and plot its magnitude and phase responses using MATLAB.

7.28 Let  $H_{LP}(z)$  denote the transfer function of an ideal real coefficient lowpass filter having a cutoff frequency of  $\omega_p$ , with  $\omega_p < \pi/2$ . Consider the complex coefficient transfer function  $H_{LP}(e^{j\omega_0}z)$ , where  $\omega_p < \omega_0 < \pi - \omega_p$ . Sketch its magnitude response for  $-\pi \leq \omega \leq \pi$ . What type of filter does it represent? Now consider the transfer function  $G(z) = H_{LP}(e^{j\omega_0}z) + H_{LP}(e^{-j\omega_0}z)$ . Sketch its magnitude response for  $-\pi \leq \omega \leq \pi$ . Show that  $G(z)$  is a real-coefficient bandpass filter with a passband centered at  $\omega_0$ . Determine the width of its passband in terms of  $\omega_p$  and its impulse response  $g[n]$  in terms of the impulse response  $h_{LP}[n]$  of the parent lowpass filter.

7.61 Let  $H_1(z)$ ,  $H_2(z)$ ,  $H_3(z)$ , and  $H_4(z)$  be, respectively, Type 1, Type 2, Type 3, and Type 4 linear-phase FIR filters. Are the following filters (composed of a cascade of the above filters) linear phase? If they are, what are their types?

- (a)  $G_a(z) = H_1(z)H_1(z)$ ,    (b)  $G_b(z) = H_1(z)H_2(z)$ ,    (c)  $G_c(z) = H_1(z)H_3(z)$ ,  
 (d)  $G_d(z) = H_1(z)H_4(z)$ ,    (e)  $G_e(z) = H_2(z)H_2(z)$ ,    (f)  $G_f(z) = H_3(z)H_3(z)$ ,  
 (g)  $G_g(z) = H_4(z)H_4(z)$ ,    (h)  $G_h(z) = H_2(z)H_3(z)$ ,    (i)  $G_i(z) = H_3(z)H_4(z)$ .

M 7.12 Design a stable second-order IIR notch filter with a center frequency at  $0.6\pi$  and a 3-dB bandwidth of  $0.2\pi$ . Plot its gain response.