COLUMBIA UNIVERSITY Department of Electrical Engineering EE EE4703: Wireless Communications

Solution for Homework Assignment No. 3

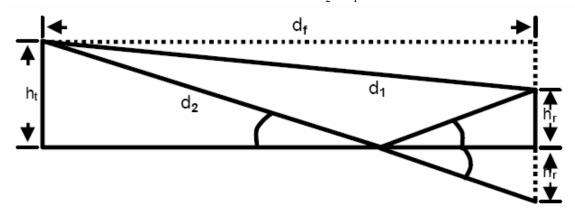
Problem 1 Rappaport, p. 169, Problem 4.15

Let df be the physical T-R separation.

d₁ is the free space propagation path length.

d2 is the ground-reflected path length.

The first Fresnel zone is defined by a df that satisfies $d_2 = d_1 + \lambda/2$.



Note that d₂ is drawn using the reflection of h_t to make the calculation easier. Also note that in this drawing $h_t > h_r$, but the same calculation applies for $h_t < h_r$.

$$\begin{aligned} d_1 &= \sqrt{d_f^2 + (h_t - h_r)^2} \\ d_2 &= \sqrt{d_f^2 + (h_t + h_r)^2} \\ \sqrt{d_f^2 + (h_t + h_r)^2} &= \sqrt{d_f^2 + (h_t - h_r)^2} + \frac{\lambda}{2} \\ d_f^2 &+ (h_t + h_r)^2 = d_f^2 + (h_t - h_r)^2 + \frac{\lambda^2}{4} + \lambda \sqrt{d_f^2 + (h_t - h_r)^2} \end{aligned}$$

$$2h_{t}h_{r} = -2h_{t}h_{r} + \frac{\mathbf{l}^{2}}{4} + \mathbf{I}\sqrt{d_{f}^{2} + (h_{t} - h_{r})^{2}}$$

$$d_{f}^{2} + (h_{t} - h_{r})^{2} = \frac{1}{\mathbf{I}^{2}} \left(4h_{t}h_{r} - \frac{\mathbf{l}^{2}}{4}\right)^{2}$$

$$d_{f}^{2} = \frac{1}{\mathbf{I}^{2}} \left(16h_{t}^{2}h_{r}^{2} - 2h_{t}h_{r}\mathbf{I}^{2} + \frac{\mathbf{l}^{4}}{16}\right) - (h_{t}^{2} - 2h_{t}h_{r} + h_{r}^{2})$$

$$d_{f}^{2} = \frac{1}{\mathbf{I}^{2}} \left(16h_{t}^{2}h_{r}^{2} + \frac{\mathbf{l}^{4}}{16} - h_{t}^{2}\mathbf{I}^{2} - h_{r}^{2}\mathbf{I}^{2}\right) = \frac{1}{\mathbf{I}^{2}} \left(\frac{\mathbf{l}^{2}}{4} - 4h_{t}^{2}\right) \left(\frac{\mathbf{l}^{2}}{4} - 4h_{r}^{2}\right)$$

$$d_{f} = \frac{\sqrt{(\mathbf{l}^{2} - 16h_{t}^{2})(\mathbf{l}^{2} - 16h_{r}^{2})}}{4\mathbf{I}}$$

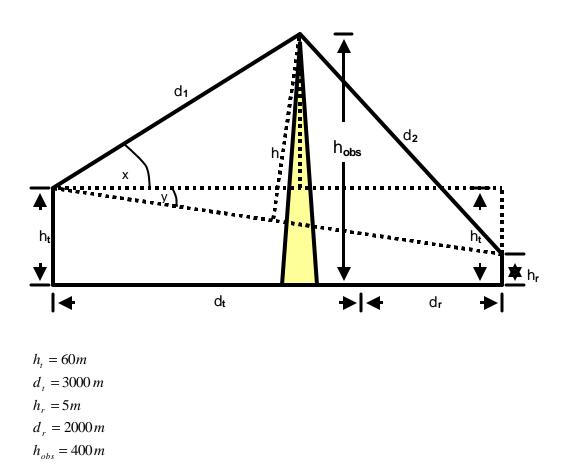
Had we used the parallel ray approximation, we would have received:

$$S = d_2 - d_1 = \frac{2h_t h_r}{d_f} = \frac{1}{2}$$
$$d_f = \frac{4h_t h_r}{1}$$

Which is correct only if the carrier wavelength is very small in comparison with the receiver and transmitter heights.

Rappaport, p. 170, Problem 4.19

 $P_{t} = 10 w = 40 dBm$ $G_{t} = 10 dB$ $G_{r} = 3 dB$ L = 1 dB @ 900 MHz f = 900 Mhz $\therefore \mathbf{l} = \frac{c}{f} = \frac{1}{3}m$



First, let's compute the free-space model:

$$PL = \frac{I^2}{[4p(d_t + d_r)]^2} = 2.81 \cdot 10^{-11} = -105.5 dB$$
$$P_{rec,dBm} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} = 40 - 105.5 + 10 + 3 - 1 = -53.5 dBm = 4.47 \cdot 10^{-9} w$$

We should extract h from the geometry of the problem

$$d_{1} = \sqrt{d_{t}^{2} + (h_{obs} - h_{t})^{2}} = 3019m$$

$$x = \arctan(\frac{h_{obs} - h_{t}}{d_{t}}) = 6.47^{\circ}$$

$$y = \arctan(\frac{h_{t} - h_{r}}{d_{t} + d_{r}}) = 0.63^{\circ}$$

$$h = d_{1} \cdot \sin(x + y) = 373.15m$$

$$v = h \cdot \sqrt{\frac{2}{I} \left(\frac{1}{d_{1}} + \frac{1}{d_{2}}\right)} = 26.39$$

$$G_{d,dB} = 20 \log\left(\frac{0.225}{v}\right) = -41.39dB$$

And then, the received power is:

$$P_{rec} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} + G_{d,dB} = -94.89 \, dBm = 3.24 \cdot 10^{-13} \, w$$

The received power is $3.24 \cdot 10^{-13}$ watts. Without the obstacle, it would have been $4.47 \cdot 10^{-9}$ watts.

The path loss caused by diffraction is 41.39dB.

Rappaport, p. 170, Problem 4.20

(a)

$$f = 50 Mhz$$

$$\therefore \mathbf{l} = \frac{c}{f} = 6m$$

$$PL = \frac{\mathbf{l}^2}{[4\mathbf{p}(d_t + d_r)]^2} = 9.12 \cdot 10^{-9} = -80.4 dB$$

Without diffraction:

$$P_{rec,dBm} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} = -28.4 dBm = 1.45 \cdot 10^{-6} w$$

And with diffraction:

$$v = h \cdot \sqrt{\frac{2}{I} \left(\frac{1}{d_1} + \frac{1}{d_2}\right)} = 6.22$$

$$G_{d,dB} = 20 \log \left(\frac{0.225}{v}\right) = -28.83 dB$$

$$P_{rec,dBm} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} + G_{d,dB} = -57.23 dBm = 1.89 \cdot 10^{-9} w$$

The received power is $1.89 \cdot 10^{-9}$ watts.

Without the obstacle, it would have been $1.45 \cdot 10^{-6}$ watts.

The path loss caused by diffraction is 28.83dB.

(b)

$$f = 1900 Mhz$$

$$\therefore \mathbf{l} = \frac{c}{f} = 0.158m$$

$$PL = \frac{\mathbf{l}^2}{[4\mathbf{p}(d_t + d_r)]^2} = 6.24 \cdot 10^{-12} = -112.05 dB$$

Without diffraction:

 $P_{rec,dBm} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} = -60.05 dBm = 9.89^{-10} w$

And with diffraction:

$$v = h \cdot \sqrt{\frac{2}{I} \left(\frac{1}{d_1} + \frac{1}{d_2}\right)} = 38.32$$

$$G_{d,dB} = 20 \log \left(\frac{0.225}{v}\right) = -44.62 \, dB$$

$$P_{rec,dBm} = P_{tr,dBm} + PL + G_{t,dB} + G_{r,dB} - L_{dB} + G_{d,dB} = -104.67 \, dBm = 3.41 \cdot 10^{-14} \, w$$

The received power is $3.41 \cdot 10^{-14}$ watts. Without the obstacle, it would have been $9.89 \cdot 10^{-10}$ watts.

The path loss caused by diffraction is 44.62 dB.

Problem 4 Rappaport, p. 170, Problem 4.22

 $P_0 = 1mw = 0dBm$ $d_0 = 1m$ d = 10m $\overline{n} = 3.5$

We require 10% of the measurements to be stronger than -25dBm. i.e. $Pr\{P_{rec} > -25dBm\} = 0.9$

We should first find the mean path loss and then look it up in the Q-function table.

$$\overline{P_{rec}} = P_0 - 10 \cdot \overline{n} \cdot \log\left(\frac{d}{d_0}\right) = -35dB$$
$$Q\left(\frac{X - \mathbf{m}}{\mathbf{s}}\right) = Q\left(\frac{(-35) - (-25)}{\mathbf{s}}\right) = Q\left(\frac{-10}{\mathbf{s}}\right) = 0.9$$
$$Q\left(\frac{10}{\mathbf{s}}\right) = 0.1 \rightarrow \frac{10}{\mathbf{s}} = 1.28$$
$$\mathbf{s} = 7.81dB$$

The standard deviation would be 7.81 dB

Rappaport, p. 172, Problem 4.27

 $P_{tr} = 15w = 41.76dBm$ $G_{t} = 12dB = 15.84$ $G_{r} = 3db = 2$ $BW_{r} = 30kHz$ $F_{rec} = 8dB$ $f_{c} = 1800 MHz$ $\therefore \mathbf{l} = 0.167m$ n = 4 $\mathbf{s} = 8dB$ $d_{0} = 1000m$ We require 20 dB SNR 95% percent of the time i.e. Pr{SNR > 20dB} = 0.95
What is the maximum T-R separation that will ensure that SNR?

We first compute the noise threshold, and then the required received power. Thermal noise spectral density at room temperature:

$$KT = -204 \, dB \frac{w}{Hz} = 3.98 \cdot 10^{-21} \frac{w}{Hz}$$

So the thermal noise at the receives is

$$KT \cdot BW_r = 1.194 \cdot 10^{-16} = -129.2dBm$$

And the noise floor is:

$$F_{rec} + KT \cdot BW_r = -121.2dBm$$

Regarding the SNR:

$$\Pr{SNR > 20dB} = Q\left(\frac{20 - \overline{SNR}}{8}\right) = 0.95$$
$$Q\left(\frac{\overline{SNR} - 20}{8}\right) = 1 - 0.95 = 0.05$$
$$\frac{\overline{SNR} - 20}{8} = 1.645$$
$$\overline{SNR} = 33.16dB$$

The required received power will be:

 $P_{rec,\min} = -121.2 + 33.16 = -88.04 dBm$

The Power at d₀:

$$P_0 = \frac{P_{tr}G_tG_r\mathbf{I}^2}{(4\mathbf{p}d_0)^2} = 8.39 \cdot 10^{-8} w = -40.76 \, dBm$$

$$P_{rec,\min} = P_0 - 10 n \cdot \log\left(\frac{d_{\max}}{d_0}\right)$$
$$\therefore d_{\max} = d_0 \cdot 10^{\frac{P_{0,dBm} - P_{rec,\min}}{10n}} = 15.2 km$$

Rappaport, p. 172, Problem 4.28

$$SNR_{req} = 25 dB$$

$$f_c = 900 MHz$$

$$\therefore \mathbf{l} = 0.333m$$

$$EIRP = P_{tr}G_t = 100W = 50 dBm$$

$$G_r = 0 dB = 1$$

$$AMPS \rightarrow BW = 30 KHz$$

$$F = 10 dB$$

$$d = 10 km$$

$$\overline{n} = 4$$

$$\mathbf{s} = 8 dB$$

$$d_0 = 1 km$$

We will find P_0 , and then the average received power.

$$P_0 = \frac{P_{tr}G_tG_r\mathbf{l}^2}{(4\mathbf{p}d_0)^2} = \frac{EIRP \cdot G_r\mathbf{l}^2}{(4\mathbf{p}d_0)^2} = 7.02 \cdot 10^{-8} w = -41.54 dBm$$

$$\overline{P_{rec}} = P_0 - 10\overline{n} \cdot \log\left(\frac{d}{d_0}\right) = -81.54 dBm$$

The thermal noise, just like in problem 4.27, is:

$$KT \cdot BW_r = 1.194 \cdot 10^{-16} = -129.2dBm$$

And the noise floor

 $NF = F_{rec} + KT \cdot BW_r = -119.2 dBm$

The mean SNR:

$$\overline{SNR} = \overline{P_{rec}} - NF = 37.66 \, dB$$
$$\Pr\{SNR > 25 \, dB\} = Q\left(\frac{25 \, dB - \overline{SNR}}{s}\right) = Q(-1.58) = 1 - Q(1.58) = 0.943$$

The required SNR will be achieved 94.3% of the time.